

FREQUENCY RESPONSE ANALYSIS

Introduction

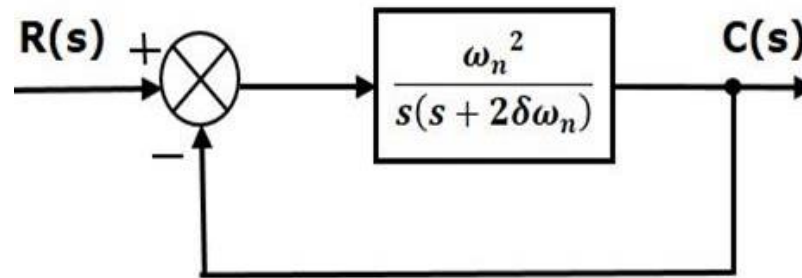
The steady state response of a system for an input sinusoidal signal is known as the **frequency response**

Consider a linear system with a sinusoidal input

$$r(t) = A \sin \omega t$$

- Under steady-state, the system output as well as the signals at all other points in the system are sinusoidal.
- The steady-state output is $c(t) = B \sin (\omega t + \phi)$
- The magnitude and phase relationship between the sinusoidal input and the steady-state of a system is termed as Frequency Response.

- In LTI systems, the frequency response is independent of amplitude and phase of the input signal.
- The analysis in frequency domain is easy than in time domain.
- The frequency response is evaluated from the sinusoidal transfer function by replacing s by $j\omega$ in the system transfer function $T(s)$
- The transfer function $T(j\omega)$ has both magnitude and phase angle.
- The characteristics are conveniently represented by graphical plots.



Consider the transfer function of the second order closed control system as

$$T(s) = \frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\delta\omega_n s + \omega_n^2}$$

Substitute, $s = j\omega$ in the above equation.

$$\begin{aligned} T(j\omega) &= \frac{\omega_n^2}{(j\omega)^2 + 2\delta\omega_n(j\omega) + \omega_n^2} \\ \Rightarrow T(j\omega) &= \frac{\omega_n^2}{-\omega^2 + 2j\delta\omega\omega_n + \omega_n^2} = \frac{\omega_n^2}{\omega_n^2 \left(1 - \frac{\omega^2}{\omega_n^2} + \frac{2j\delta\omega}{\omega_n}\right)} \\ \Rightarrow T(j\omega) &= \frac{1}{\left(1 - \frac{\omega^2}{\omega_n^2}\right) + j\left(\frac{2\delta\omega}{\omega_n}\right)} \end{aligned}$$

Let, $\frac{\omega}{\omega_n} = u$ Substitute this value in the above equation.

$$T(j\omega) = \frac{1}{(1 - u^2) + j(2\delta u)}$$

Magnitude of $T(j\omega)$ is -

$$M = |T(j\omega)| = \frac{1}{\sqrt{(1-u^2)^2 + (2\delta u)^2}}$$

Phase of $T(j\omega)$ is -

$$\phi = \angle T(j\omega) = -\tan^{-1} \left(\frac{2\delta u}{1-u^2} \right)$$

The loop transfer function of a system is given by $G(s) = \frac{100(s+2)}{s(s+1)(s+5)}$

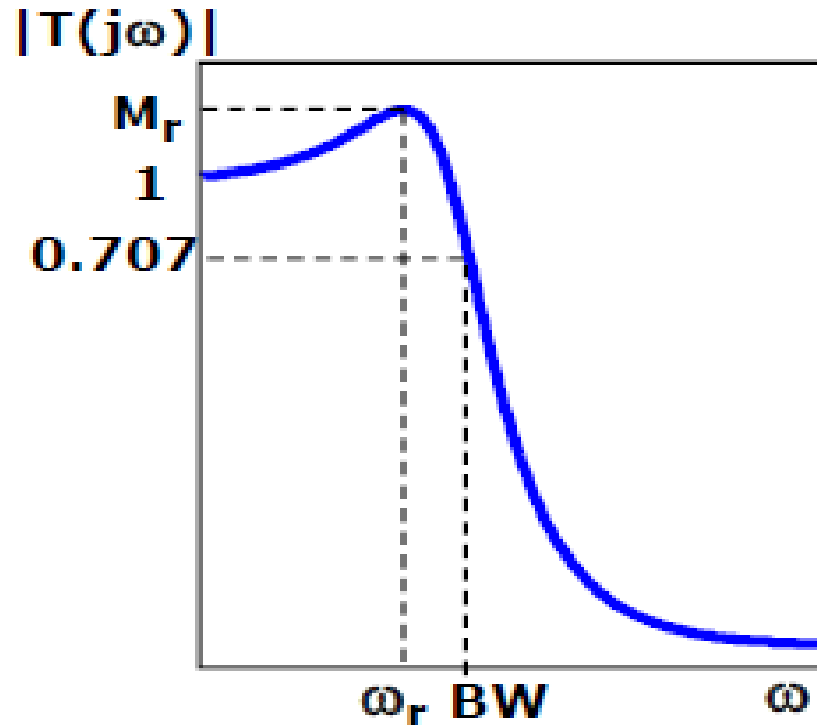
Find the magnitude and phase at 10 rad/sec.

$$M = A + jB = \sqrt{A^2 + B^2}$$

$$\phi = \tan^{-1} B/A$$

Frequency Domain Specifications

- Resonant peak
- Resonant frequency
- Bandwidth.



Resonant Frequency

Resonant frequency, ω_r :

This is the frequency at which the resonant peak is obtained.

Differentiate M with respect to u .

$$\begin{aligned}\frac{dM}{du} &= -\frac{1}{2} [(1 - u^2)^2 + (2\delta u)^2]^{-\frac{3}{2}} [2(1 - u^2)(-2u) + 2(2\delta u)(2\delta)] \\ \Rightarrow \frac{dM}{du} &= -\frac{1}{2} [(1 - u^2)^2 + (2\delta u)^2]^{-\frac{3}{2}} [4u(u^2 - 1 + 2\delta^2)]\end{aligned}$$

Substitute, $u = u_r$ and $\frac{dM}{du} = 0$ in the above equation.

$$\begin{aligned}0 &= -\frac{1}{2} [(1 - u_r^2)^2 + (2\delta u_r)^2]^{-\frac{3}{2}} [4u_r(u_r^2 - 1 + 2\delta^2)] \\ \Rightarrow 4u_r(u_r^2 - 1 + 2\delta^2) &= 0 \\ \Rightarrow u_r^2 - 1 + 2\delta^2 &= 0 \\ \Rightarrow u_r^2 &= 1 - 2\delta^2\end{aligned}$$

$$\Rightarrow u_r = \sqrt{1 - 2\delta^2}$$

Substitute, $u_r = \frac{\omega_r}{\omega_n}$ in the above equation.

$$\begin{aligned}\frac{\omega_r}{\omega_n} &= \sqrt{1 - 2\delta^2} \\ \Rightarrow \omega_r &= \omega_n \sqrt{1 - 2\delta^2}\end{aligned}$$

Resonant Peak

Resonant Peak

It is the peak (maximum) value of the magnitude of $T(j\omega)$. It is denoted by M_r .

At $u=u_r$, the Magnitude of $T(j\omega)$ is -

$$M_r = \frac{1}{\sqrt{(1 - u_r^2)^2 + (2\delta u_r)^2}}$$

Substitute, $u_r = \sqrt{1 - 2\delta^2}$ and $1 - u_r^2 = 2\delta^2$ in the above equation.

$$M_r = \frac{1}{\sqrt{(2\delta^2)^2 + (2\delta\sqrt{1 - 2\delta^2})^2}}$$

$$\Rightarrow M_r = \frac{1}{2\delta\sqrt{1 - \delta^2}}$$

Bandwidth

Bandwidth

It is the range of frequencies over which, the magnitude of $T(j\omega)$ drops to 70.7% from its zero frequency value.

At $\omega=0$, the value of u will be zero.

Substitute, $u=0$ in M .

$$M = \frac{1}{\sqrt{(1 - 0^2)^2 + (2\delta(0))^2}} = 1$$

Therefore, the magnitude of $T(j\omega)$ is one at $\omega=0$

At 3-dB frequency, the magnitude of $T(j\omega)$ will be 70.7% of magnitude of $T(j\omega)$ at $\omega=0$

i.e., at $\omega = \omega_B$, $M = 0.707(1) = \frac{1}{\sqrt{2}}$

$$\Rightarrow M = \frac{1}{\sqrt{2}} = \frac{1}{\sqrt{(1 - u_b^2)^2 + (2\delta u_b)^2}}$$

$$\Rightarrow 2 = (1 - u_b^2)^2 + (2\delta)^2 u_b^2$$

$$\text{Let, } u_b^2 = x$$

$$\Rightarrow 2 = (1 - x)^2 + (2\delta)^2 x$$

$$\Rightarrow x^2 + (4\delta^2 - 2)x - 1 = 0$$

$$\Rightarrow x = \frac{-(4\delta^2 - 2) \pm \sqrt{(4\delta^2 - 2)^2 + 4}}{2}$$

Consider only the positive value of x.

$$x = 1 - 2\delta^2 + \sqrt{(2\delta^2 - 1)^2 + 1}$$

$$\Rightarrow x = 1 - 2\delta^2 + \sqrt{(2 - 4\delta^2 + 4\delta^4)}$$

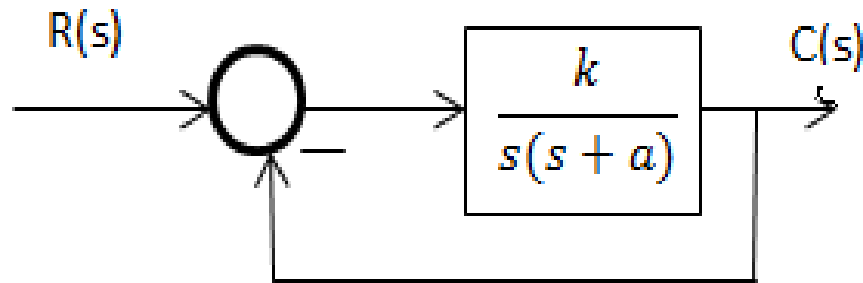
$$\text{Substitute, } x = u_b^2 = \frac{\omega_b^2}{\omega_n^2}$$

$$\frac{\omega_b^2}{\omega_n^2} = 1 - 2\delta^2 + \sqrt{(2 - 4\delta^2 + 4\delta^4)}$$

$$\Rightarrow \omega_b = \omega_n \sqrt{1 - 2\delta^2 + \sqrt{(2 - 4\delta^2 + 4\delta^4)}}$$

Problem 1

Consider the system as shown in figure,



(a) Find the value 'k' and 'a' to satisfy the following frequency domain specifications:

$$M_r = 1.04$$

$$\omega_r = 11.55 \text{ rad/sec}$$

(b) For this value of k and a , calculate settling time and bandwidth of the system

Solution

$$\frac{C(s)}{R(s)} = \frac{k}{s^2 + as + k};$$

$$\omega_n^2 = k \quad 2\zeta\omega_n = a$$

$$M_r = 1.04$$

$$\omega_r = 11.55 \text{ rad/sec}$$

$$M_r = \frac{1}{2\sqrt{1-2\zeta^2}}$$

$$\omega_r = \omega_n \sqrt{1-2\zeta^2}$$

$$\zeta = 0.6, 0.8$$

$$\omega_n = 21.8 \text{ rad/sec}$$

$$k = 476$$

$$a = 26$$

Problem 2

Unit- step response data of a second-order system is given below. Obtain the corresponding frequency response specifications for the system

| | | | | | | | | | | | |
|--------------------|---|------|------|------|------|------|------|------|------|------|------|
| t (sec) | 0 | 0.05 | 0.10 | 0.15 | 0.20 | 0.25 | 0.30 | 0.35 | 0.40 | 0.45 | 0.50 |
| C(t) | 0 | 0.25 | 0.8 | 1.08 | 1.12 | 1.02 | 0.98 | 0.98 | 1.0 | 1.0 | 1.0 |

Solution

From the table, $t_p = 0.2$

$$M_p = 0.12$$

Find ζ and ω_n

Determine M_r , ω_r and BW.

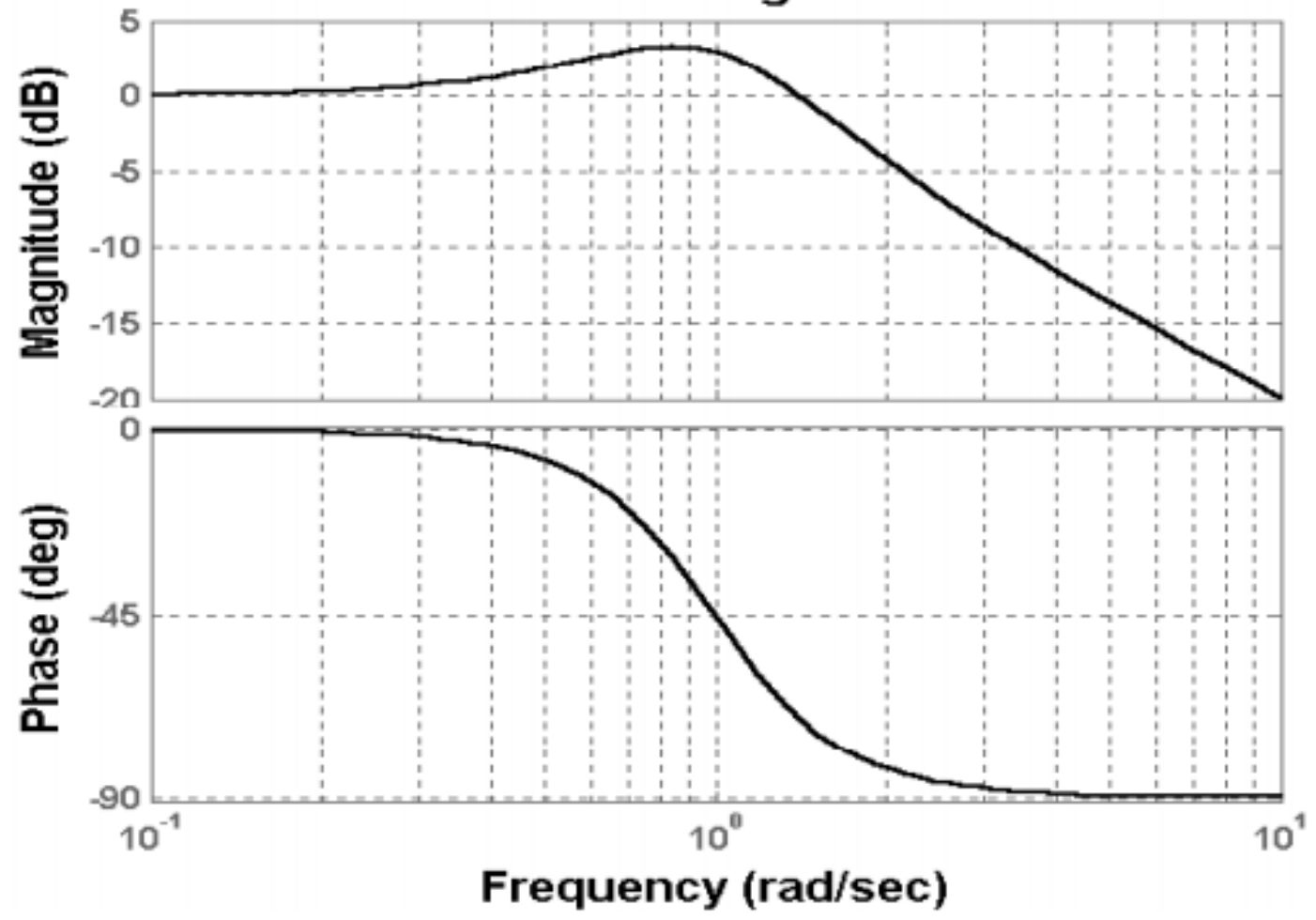
Frequency Domain Plots

1. Bode plot or Logarithmic plot
2. Nyquist plot

BODE PLOT

- The Bode plot of a transfer function is a useful graphical tool for the analysis and design of linear control systems in the frequency domain
- The Bode plot consists of two plots drawn on semi-logarithmic paper.
 1. Magnitude of the frequency response in decibels, i.e., $20 \log |G(j\omega)|$ on a linear scale versus frequency on a logarithmic scale.
 2. Phase of the frequency response function on a linear scale versus frequency on a logarithmic scale.

Bode Diagram



Basic factors

Consider the following general transfer function

$$G(s) = \frac{k(1+Tas)(1+Tbs).....}{s^r(1+T_1s)(1+T_2s).....(s^2+2\zeta\omega_n s+\omega_n^2)}$$

$$G(j\omega) = \frac{k(1+j\omega T_a)(1+j\omega T_b).....}{(j\omega)^r(1+j\omega T_1)(1+j\omega T_2)..\left[1+j2\zeta\left(\frac{\omega}{\omega_n}\right)+\left(\frac{j\omega}{\omega_n}\right)^2\right]..}$$

1. Gain k (Constant Term)
2. Integral or Derivative factors $(j\omega)^{\pm 1}$
(Poles or zeros at origin)
3. First-order factors $(1+j\omega T)^{\pm 1}$
(poles or zeros not at origin)
4. Quadratic factors $[1+2\zeta(j\omega/\omega_n)+(j\omega/\omega_n)^2]^{\pm 1}$
(Complex poles or Complex zeros)

The magnitude of $G(j\omega) = |G(j\omega)|$

$$20\log |G(j\omega)| = 20\log |k| + 20\log |1+j\omega T_a| + 20\log |1+j\omega T_b| - \\ 20r \log \omega - 20\log |1+j\omega T_1| - 20\log |1+j\omega T_2| \dots - \\ 20\log \left| 1 + j2\zeta \left(\frac{\omega}{\omega_n} \right) - \left(\frac{\omega}{\omega_n} \right)^2 \right| \dots$$

The Phase of $G(j\omega) = \angle G(j\omega)$

$$\angle G(j\omega) = \tan^{-1}\omega T_a + \tan^{-1}\omega T_b + \dots - r(90^\circ) - \tan^{-1}\omega T_1 - \\ \tan^{-1}\omega T_2 - \dots - \tan^{-1} \left\{ \frac{2\zeta \left(\frac{\omega}{\omega_n} \right)}{1 - \left(\frac{\omega}{\omega_n} \right)^2} \right\} \dots$$

Gain K, Constant Term

- The gain factor multiplies the overall gain by a constant value for all frequencies.
- It has no effect on phase.

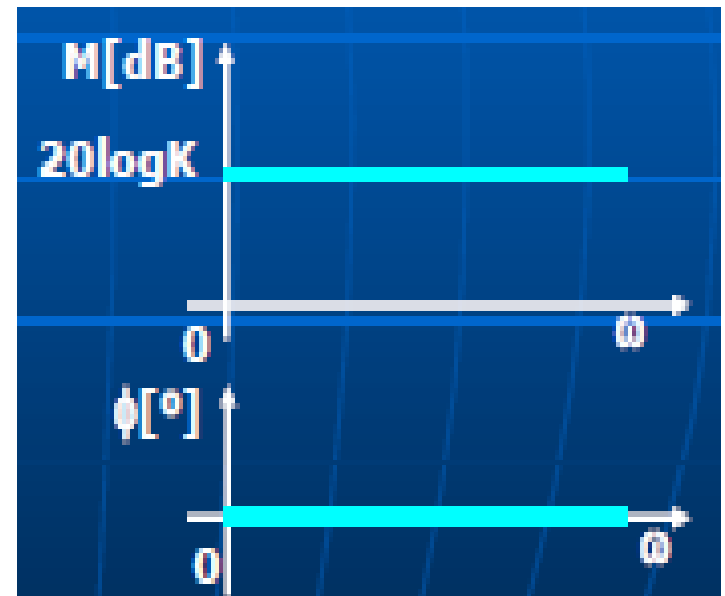
$$G(s)=K$$

$$G(j\omega)=k$$

$$M = 20\log |G(j\omega)|$$

$$M = 20\log K = \text{constant}$$

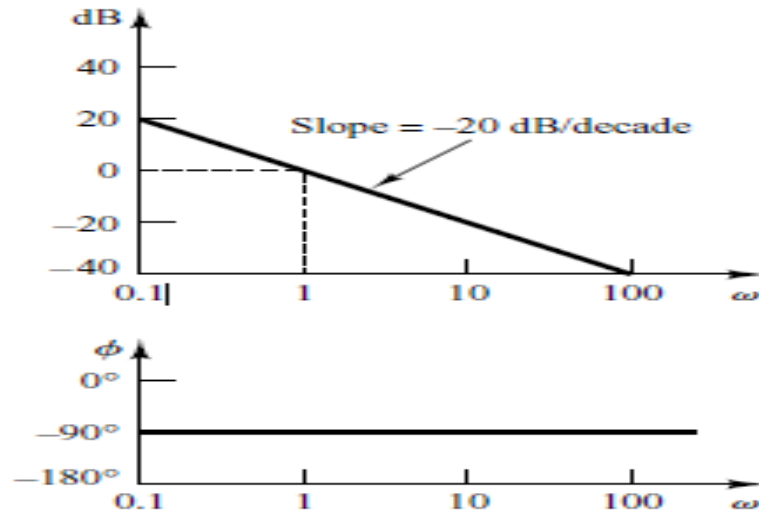
$$\phi = \angle G(j\omega) = \tan^{-1} \frac{0}{k} = 0^\circ$$



Integral Factor $1/j\omega$ – pole at origin

$$G(s) = \frac{1}{s}$$

$$G(j\omega) = \frac{1}{j\omega}$$



$$M = 20 \log \left| \frac{1}{j\omega} \right| = -20 \log \omega$$

Magnitude is a straight line with a slope of -20 dB/decade

$$\phi = -\tan^{-1} \frac{\omega}{0} = -90^\circ$$

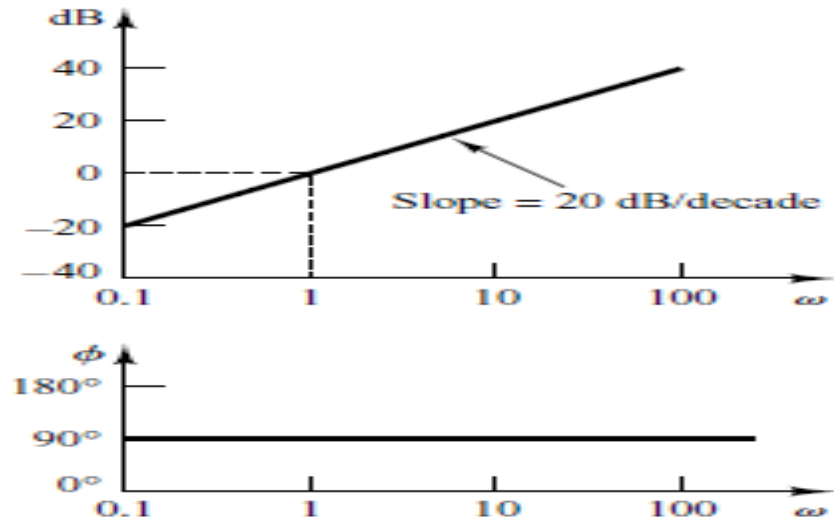
Phase is constant at -90° at all frequencies.

Derivative Factor $j\omega$ – Zero at origin

$$G(s) = s$$

$$G(j\omega) = j\omega$$

$$M = 20 \log \omega$$



Magnitude is a straight line with a slope of 20 dB/decade

$$\phi = \tan^{-1} \frac{\omega}{0} = 90^\circ$$

Phase is constant at 90° at all frequencies

Poles not at origin

$$G(s) = \frac{1}{1+Ts}$$

$$G(j\omega) = \frac{1}{1+Tj\omega}$$

$$M = 20\log \left| \frac{1}{1+Tj\omega} \right| \qquad \emptyset = -\tan^{-1} \omega T$$

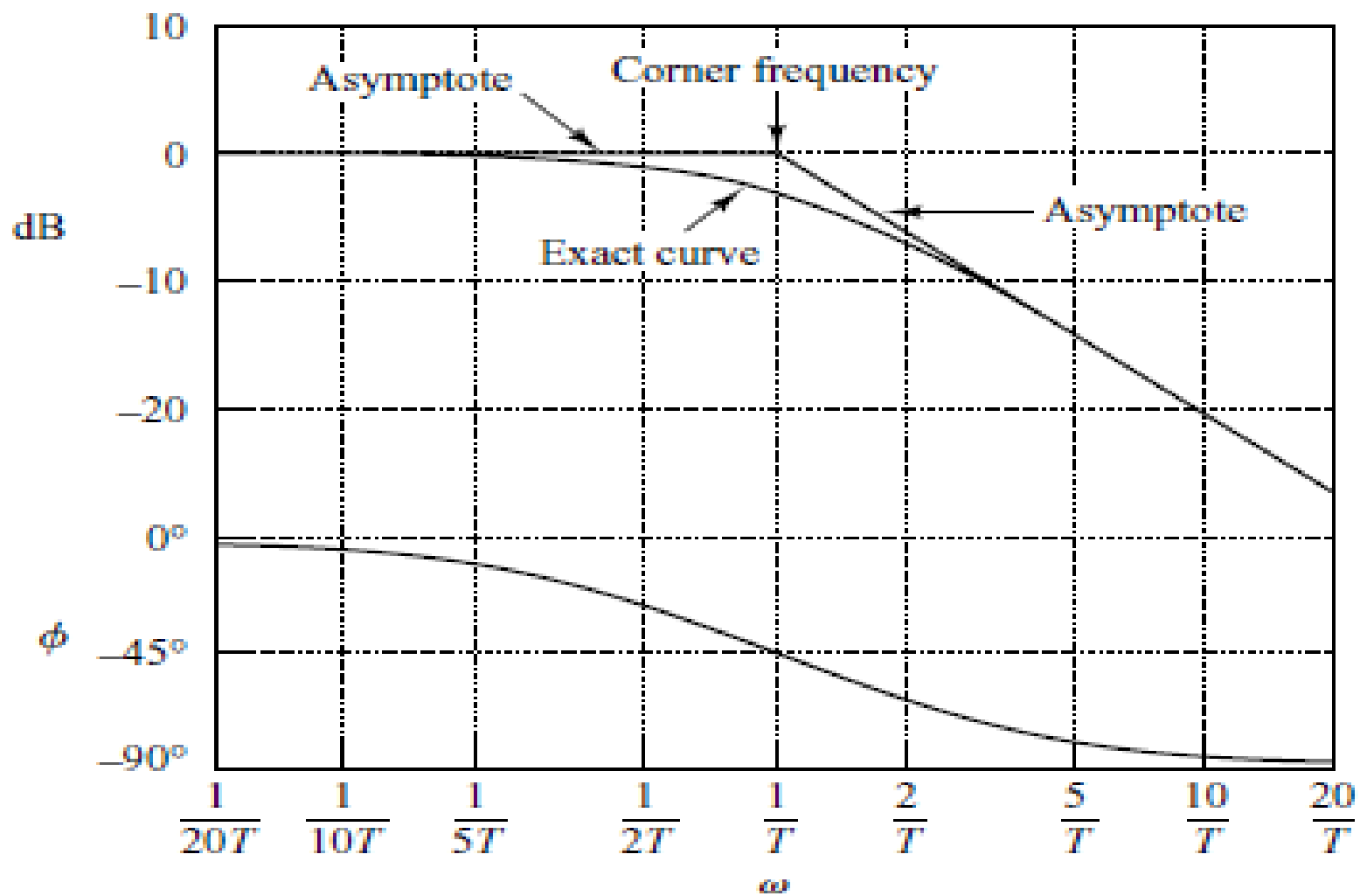
$$= -20 \log \sqrt{1+(\omega T)^2}$$

$$\omega T \ll 1, \quad M = -20\log 1 = 0 \text{ db} \qquad \emptyset = 0^\circ$$

$$\omega T = 1, \quad M = -20\log \sqrt{2} = -3 \text{ db} \qquad \emptyset = -45^\circ$$

(corner frequency)

$$\omega T \gg 1, \quad M = -20\log \omega T \qquad \emptyset = -90^\circ$$



Zeros not at origin

$$G(s) = 1 + Ts$$

$$G(j\omega) = 1 + j\omega T$$

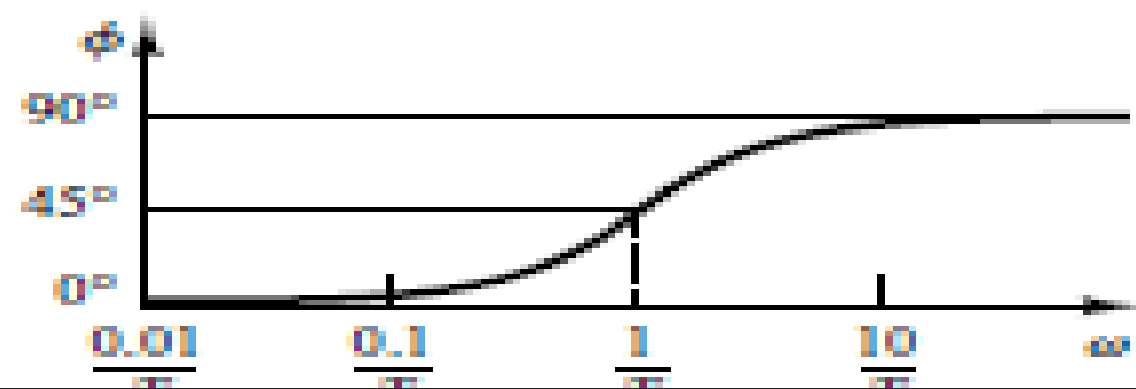
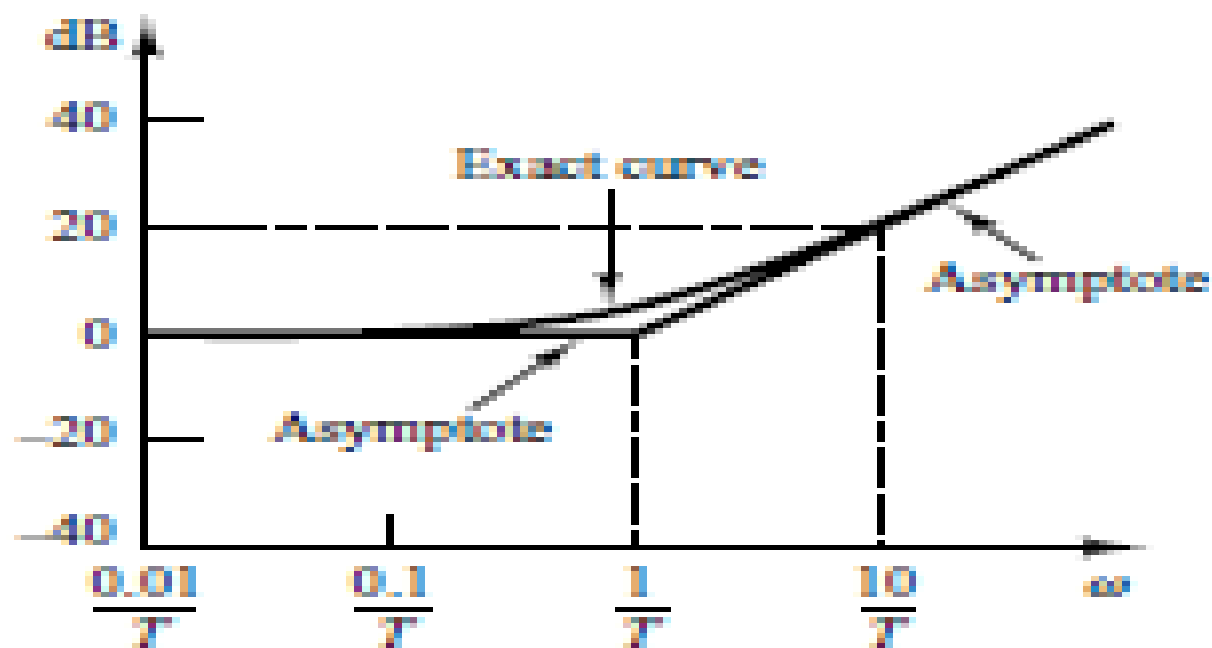
$$M = 20 \log |1 + j\omega T| \qquad \emptyset = \tan^{-1} \omega T$$
$$= 20 \log \sqrt{1 + (\omega T)^2}$$

$$\omega T \ll 1, \quad M = 20 \log 1 = 0 \text{ db} \qquad \emptyset = 0^\circ$$

$$\omega T = 1, \quad M = 20 \log \sqrt{2} = 3 \text{ db} \qquad \emptyset = 45^\circ$$

(corner frequency)

$$\omega T \gg 1, \quad M = 20 \log \omega T \qquad \emptyset = 90^\circ$$



Complex Poles

$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$s = j\omega$$

$$G(j\omega) = \frac{\omega_n^2}{(j\omega)^2 + 2\zeta\omega_n j\omega + \omega_n^2} = \frac{\omega_n^2}{\omega_n^2 \left[1 + j2\zeta\left(\frac{\omega}{\omega_n}\right) + \left(\frac{j\omega}{\omega_n}\right)^2 \right]}$$

$$= \frac{1}{\left[1 + j2\zeta\left(\frac{\omega}{\omega_n}\right) - \left(\frac{\omega}{\omega_n}\right)^2 \right]} = \frac{1}{\left[1 - \left(\frac{\omega}{\omega_n}\right)^2 + j2\zeta\left(\frac{\omega}{\omega_n}\right) \right]}$$

$$M = -20 \log \left[\sqrt{\left\{ 1 - \left(\frac{\omega}{\omega_n}\right)^2 \right\}^2 + \left\{ 2\zeta\left(\frac{\omega}{\omega_n}\right) \right\}^2} \right]$$

$$\phi = -\tan^{-1} \left\{ \frac{2\zeta\left(\frac{\omega}{\omega_n}\right)}{1 - \left(\frac{\omega}{\omega_n}\right)^2} \right\}$$

Magnitude: $M = -20\log \left[\sqrt{\left\{1 - \left(\frac{\omega}{\omega_n}\right)^2\right\}^2 + \left\{2\zeta\left(\frac{\omega}{\omega_n}\right)\right\}^2} \right]$

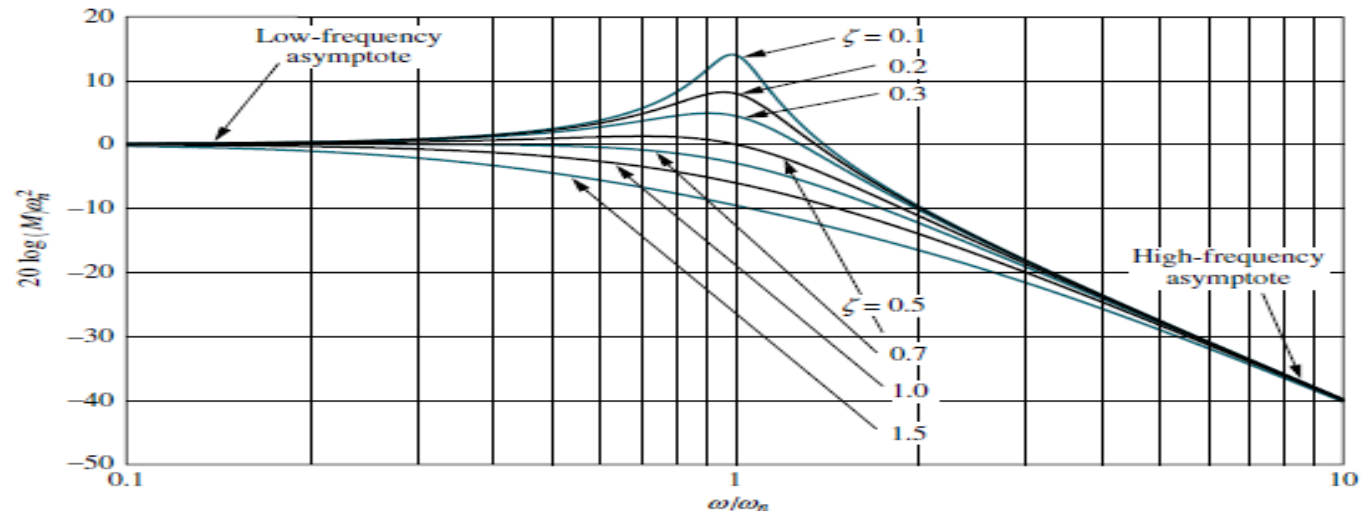
For $\zeta < 1$,

$\omega \ll \omega_n$, $M = -20\log 1 = 0$ db

$\omega = \omega_n$, $M = -20\log (2\zeta)$

(corner frequency)

$\omega \gg \omega_n$, $M = -20\log \left(\frac{\omega}{\omega_n}\right)^2 = -40\log \left(\frac{\omega}{\omega_n}\right)$

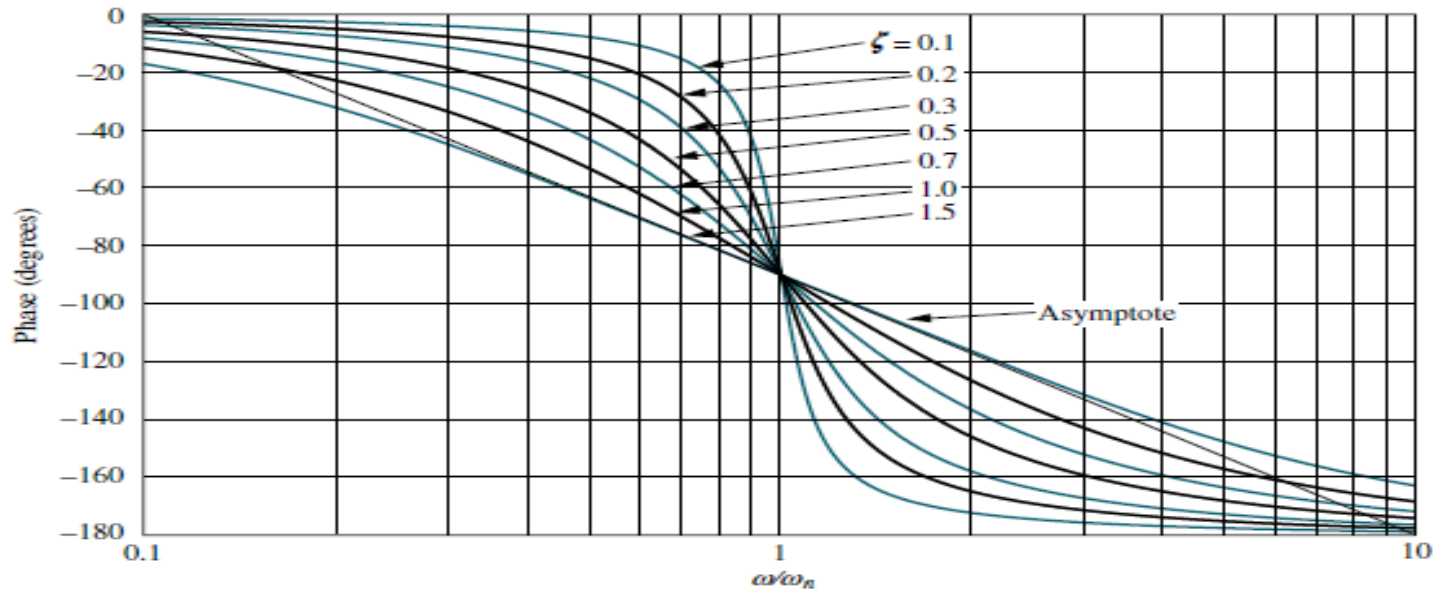


Phase: $\phi = -\tan^{-1}\left\{\frac{2\zeta\left(\frac{\omega}{\omega_n}\right)}{1-\left(\frac{\omega}{\omega_n}\right)^2}\right\}$

$\omega \ll \omega_n,$ $\phi = 0^\circ$

$\omega = \omega_n,$ (corner frequency) $\phi = -90^\circ$

$\omega \gg \omega_n,$ $\phi = -180^\circ$



Complex Zeros

$$G(s) = s^2 + 2\zeta\omega_n s + \omega_n^2$$

$$s = j\omega$$

$$G(j\omega) = [(j\omega)^2 + 2\zeta\omega_n j\omega + \omega_n^2]$$

$$= \omega_n^2 \left[1 + j2\zeta \left(\frac{\omega}{\omega_n} \right) + \left(\frac{j\omega}{\omega_n} \right)^2 \right]$$

$$= \omega_n^2 \left[1 + j2\zeta \left(\frac{\omega}{\omega_n} \right) - \left(\frac{\omega}{\omega_n} \right)^2 \right]$$

$$M = 20 \log v \left[\left\{ 1 - \left(\frac{\omega}{\omega_n} \right)^2 \right\}^2 + \left\{ 2\zeta \left(\frac{\omega}{\omega_n} \right) \right\}^2 \right]$$

$$\phi = \tan^{-1} \left\{ \frac{2\zeta \left(\frac{\omega}{\omega_n} \right)}{1 - \left(\frac{\omega}{\omega_n} \right)^2} \right\}$$

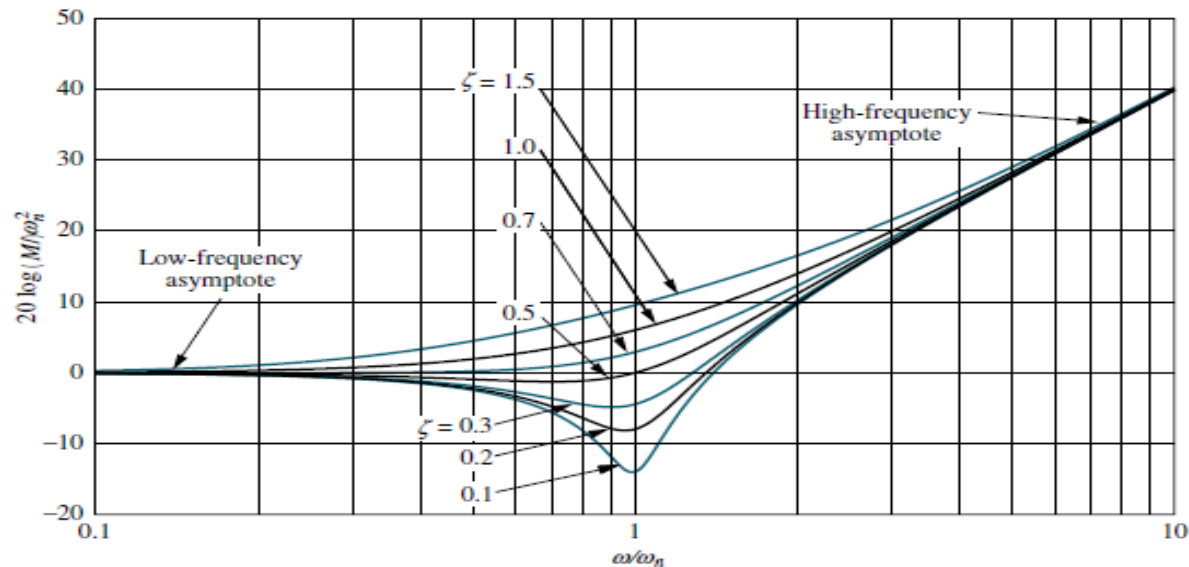
Magnitude: $M = 20 \log \left[\sqrt{\left\{ 1 - \left(\frac{\omega}{\omega_n} \right)^2 \right\}^2 + \left\{ 2\zeta \left(\frac{\omega}{\omega_n} \right) \right\}^2} \right]$

For $\zeta < 1$,

$\omega \ll \omega_n$, $M = 20 \log 1 = 0$ db

$\omega = \omega_n$, $M = 20 \log (2\zeta)$ (corner frequency)

$\omega \gg \omega_n$, $M = 20 \log \left(\frac{\omega}{\omega_n} \right)^2 = 40 \log \left(\frac{\omega}{\omega_n} \right)$

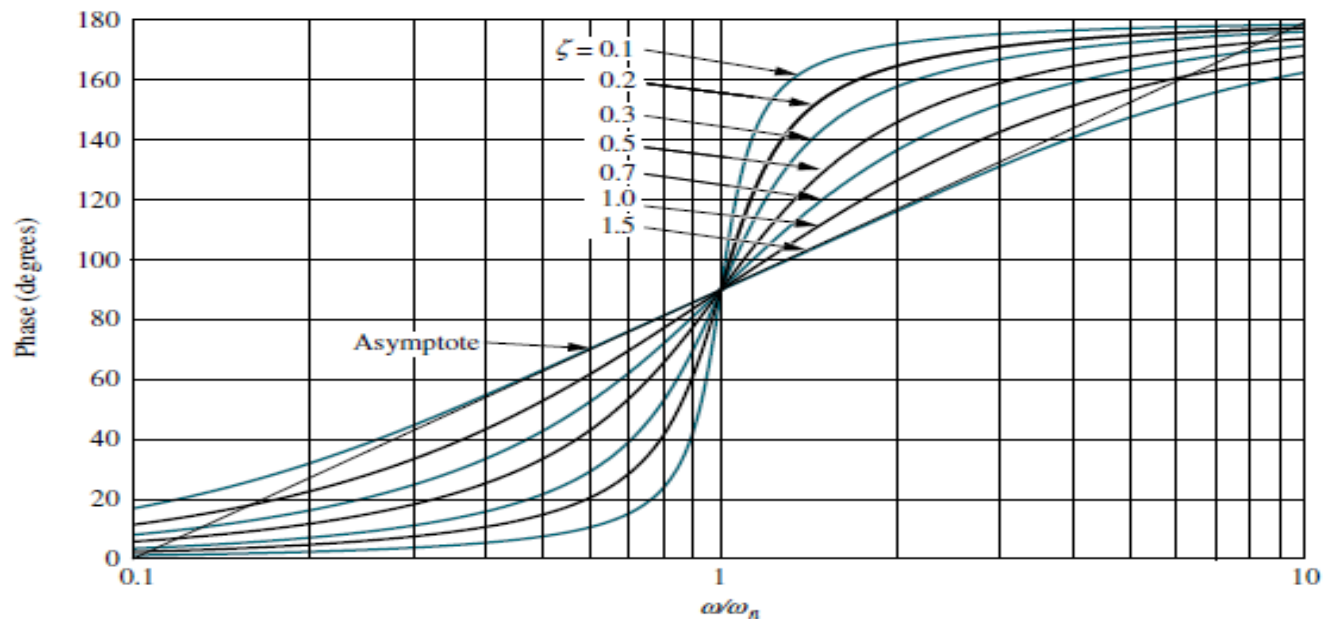


Phase: $\emptyset = \tan^{-1} \left\{ \frac{2\zeta \left(\frac{\omega}{\omega_n} \right)}{1 - \left(\frac{\omega}{\omega_n} \right)^2} \right\}$

$\omega \ll \omega_n, \emptyset = 0^\circ$

$\omega = \omega_n, \emptyset = 90^\circ$ (corner frequency)

$\omega \gg \omega_n, \emptyset = 180^\circ$



Problem 1

Draw the Bode plot for the transfer function $G(s) = \frac{10}{s(1+0.5s)(1+0.1s)}$

Solution:

$$G(s) = \frac{10}{s(1+0.5s)(1+0.1s)}$$

$$s=j\omega$$

$$G(j\omega) = \frac{10}{j\omega(1+0.5j\omega)(1+0.1j\omega)}$$

$$G(j\omega) = |G(j\omega)| \angle G(j\omega)$$

$$|G(j\omega)| = \left| \frac{10}{j\omega(1+0.5j\omega)(1+0.1j\omega)} \right|$$

$$\angle G(j\omega) = \angle \frac{10}{j\omega(1+0.5j\omega)(1+0.1j\omega)}$$

$$|G(j\omega)| = \left| \frac{10}{j\omega(1+0.5j\omega)(1+0.1j\omega)} \right|$$

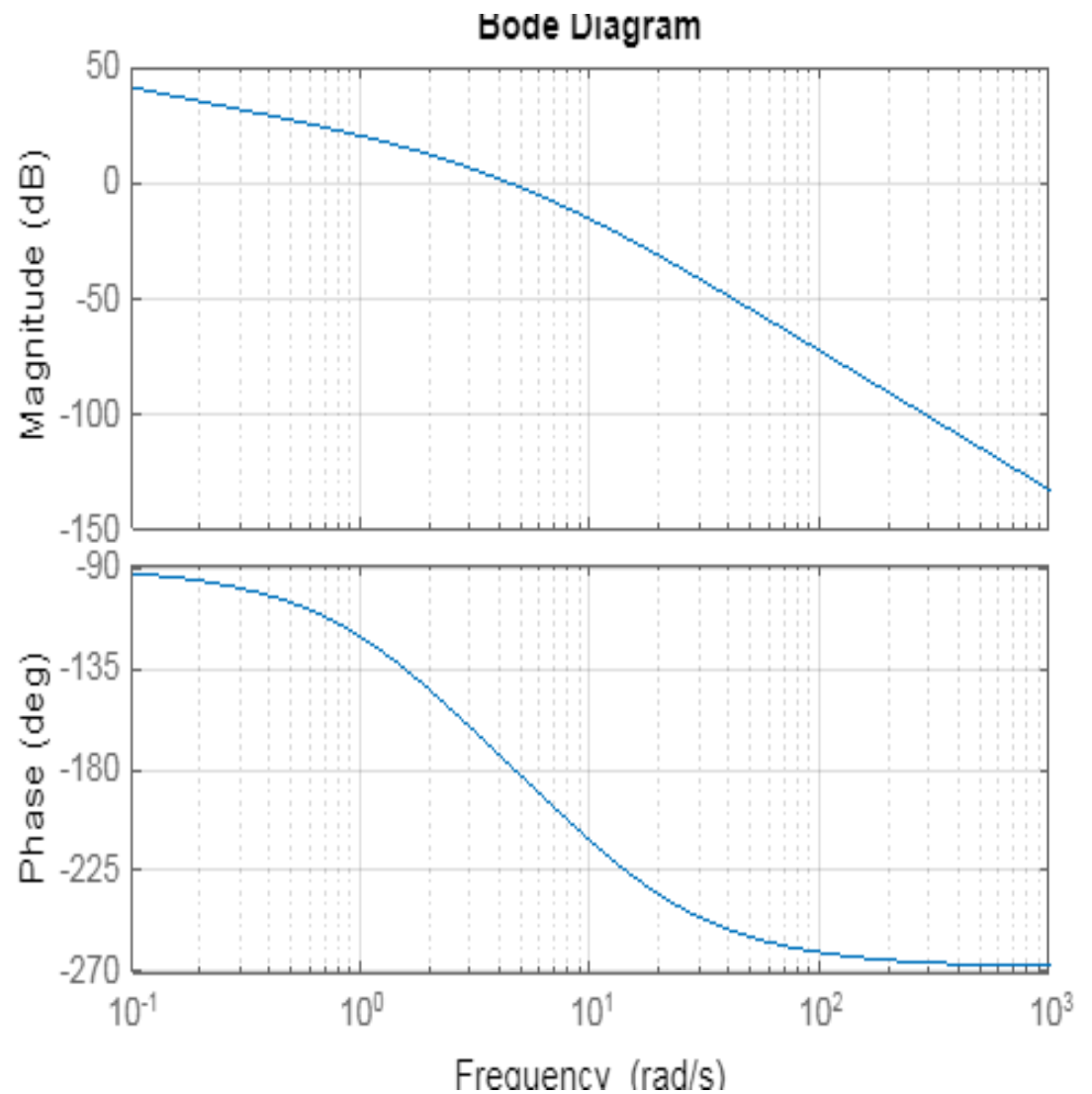
$$\text{Magnitude (db)} = 20\log |G(j\omega)|$$

$$M_{\text{db}} = 20\log 10 - 20\log \omega - 20\log \sqrt{1+(0.5\omega)^2} - 20\log \sqrt{1+(0.1\omega)^2}$$

$$\angle G(j\omega) = \angle \frac{10}{j\omega(1+0.5j\omega)(1+0.1j\omega)}$$

$$\begin{aligned} \phi &= \tan^{-1}\left(\frac{0}{10}\right) - \tan^{-1}\left(\frac{\omega}{0}\right) - \tan^{-1}\left(\frac{0.5\omega}{1}\right) - \tan^{-1}\left(\frac{0.1\omega}{1}\right) \\ &= 0 - 90 - \tan^{-1}(0.5\omega) - \tan^{-1}(0.1\omega) \end{aligned}$$

| ω | M (db) | ϕ (deg) |
|----------|-----------|-----------------|
| 0.1 | 40 | -93 |
| 1 | 19 | -122.3 |
| 2 | 11 | -146 |
| 10 | -17.16 | -213.7 |
| 100 | -74 | -263 |



Problem 2

Draw the Bode plot for the transfer function $G(s) = \frac{(s+3)}{s(s+1)(s+2)}$

$$G(s) = \frac{(s+3)}{s(s+1)(s+2)} = \frac{3(\frac{s}{3}+1)}{s(s+1)2(\frac{s}{2}+1)} = \frac{1.5(1+0.3s)}{s(s+1)(1+0.5s)}$$

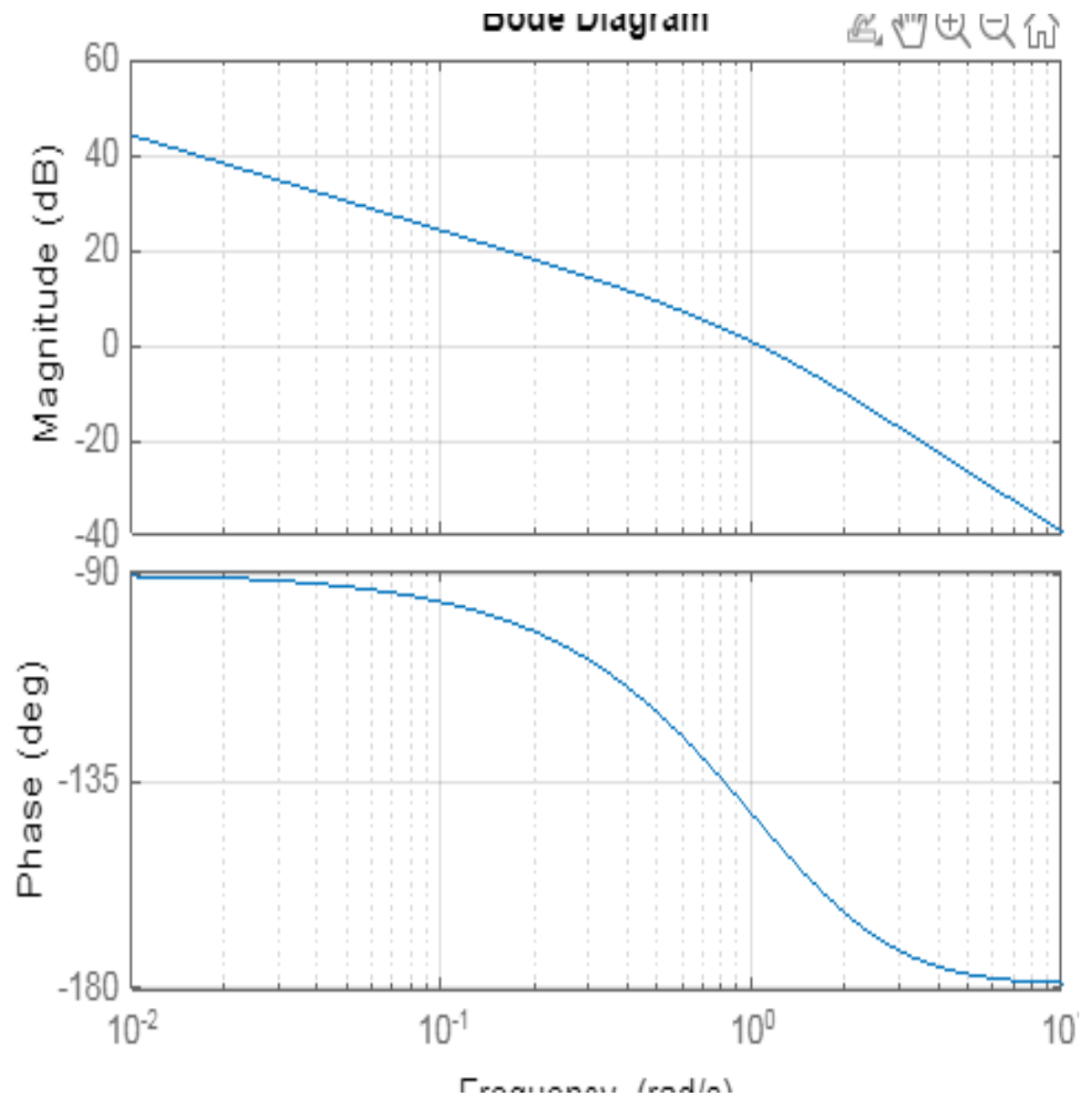
$$G(j\omega) = \frac{1.5(1+0.3j\omega)}{j\omega(j\omega+1)(1+0.5j\omega)}$$

$$M_{db} = 20\log 1.5 + 20\log \sqrt{1+(0.3\omega)^2} - 20\log \omega - 20\log \sqrt{1+(\omega)^2} - 20\log \sqrt{1+(0.5\omega)^2}$$

$$\phi = \tan^{-1}\left(\frac{0}{1.5}\right) + \tan^{-1}\left(\frac{0.3\omega}{1}\right) - \tan^{-1}\left(\frac{\omega}{0}\right) - \tan^{-1}\left(\frac{\omega}{1}\right) - \tan^{-1}\left(\frac{0.5\omega}{1}\right)$$

$$\phi = 0 + \tan^{-1}(0.3\omega) - 90 - \tan^{-1}\omega - \tan^{-1}(0.5\omega)$$

| ω | M | ϕ |
|----------|--------|--------|
| 0.1 | 23.5 | -97 |
| 1 | -0.08 | -145 |
| 2 | -11.16 | -167 |
| 3 | -18.6 | -176 |
| 10 | -40.7 | -181 |
| 100 | -81 | -180 |



Problem 3

The open loop transfer function of a unity feedback system is given by $\frac{64(s+2)}{s(s+0.5)(s^2+3.2s+64)}$. Draw the Bode plot and hence comment on stability.

$$G(s) = \frac{64(s+2)}{s(s+0.5)(s^2+3.2s+64)} = \frac{64*2 \left(\frac{s}{2}+1\right)}{s*0.5\left(\frac{s}{0.5}+1\right)*64*\left(\frac{s^2}{64}+\frac{3.2s}{64}+1\right)}$$

$$= \frac{4(0.5s+1)}{s(2s+1)(0.015s^2+0.05s+1)}$$

$$S=j\omega$$

$$G(j\omega) = \frac{4(0.5j\omega+1)}{j\omega(2j\omega+1)(0.015(j\omega)^2+0.05j\omega+1)} = \frac{4(0.5j\omega+1)}{j\omega(2j\omega+1)(1-0.015\omega^2+0.05j\omega)}$$

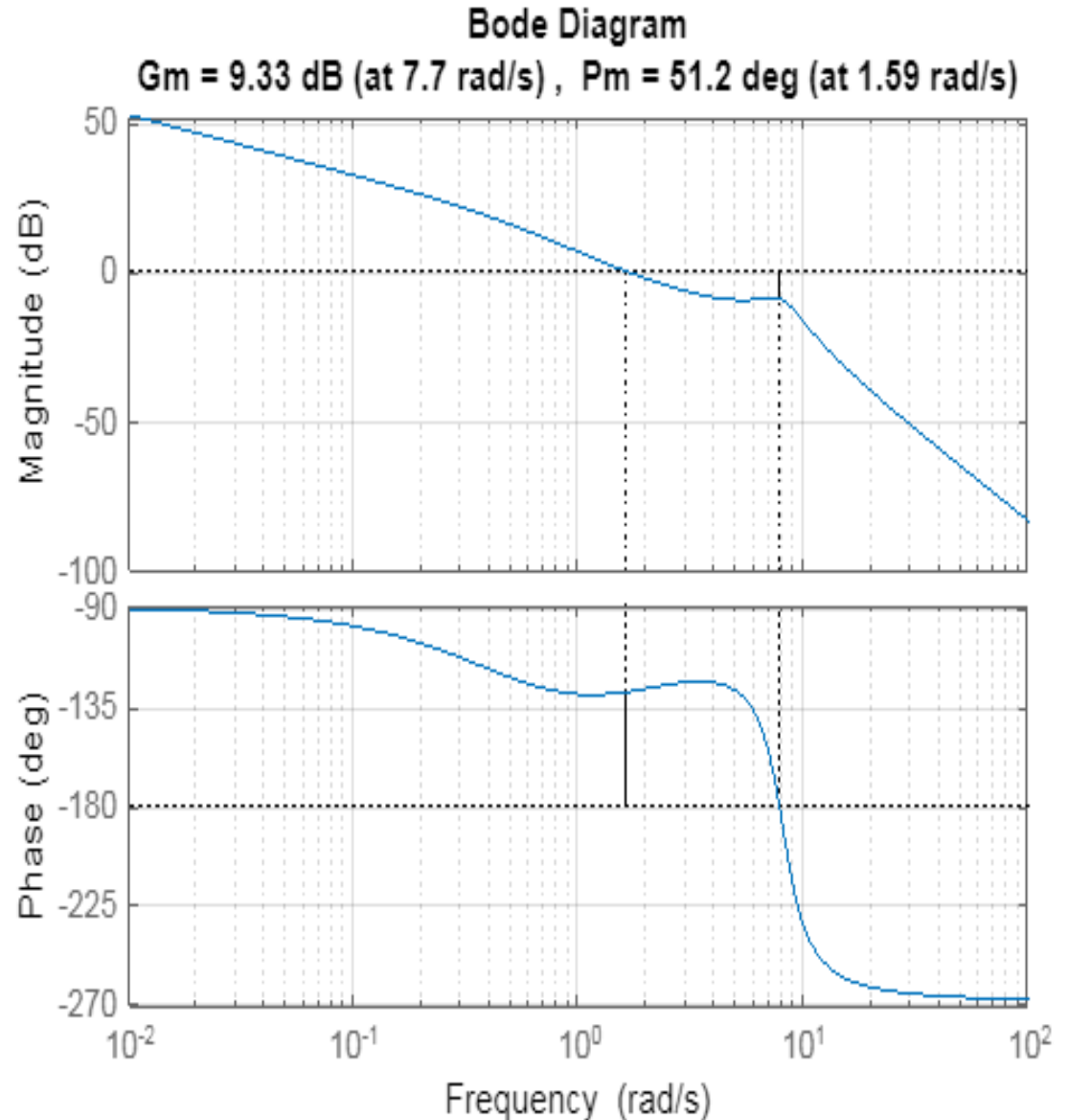
$$M_{db} = 20\log 4 + 20\log \sqrt{(1+(0.5\omega)^2)} - 20\log \omega -$$

$$20\log \sqrt{(1+(2\omega)^2)} - 20\log \sqrt{\{(1-0.015\omega^2)^2+(0.05\omega)^2\}}$$

$$\emptyset = \tan^{-1}\left(\frac{0}{4}\right) + \tan^{-1}\left(\frac{0.5\omega}{1}\right) - \tan^{-1}\left(\frac{\omega}{0}\right) - \tan^{-1}\left(\frac{2\omega}{1}\right) - \tan^{-1}\left(\frac{0.05\omega}{1-0.015\omega^2}\right)$$

$$= 0 + \tan^{-1} 0.5\omega - 90 - \tan^{-1} 2\omega - \tan^{-1}\left(\frac{0.05\omega}{1-0.015\omega^2}\right)$$

| ω | M | ϕ |
|----------|-----|----------------------|
| 0.1 | 32 | -99 |
| 0.5 | 15 | -122 |
| 1 | 6 | -130 |
| 2 | -3 | -127 |
| 3 | -7 | -124 |
| 8 | -9 | -185 |
| 10 | -17 | -53(-180) = -233 |
| 100 | -83 | -89 (-180) = -269 |



Problem 4

Draw the Bode plot for the transfer function $G(s) = \frac{10(s+3)}{s(s+2)(s^2+2s+2)}$

$$G(s) = \frac{10(s+3)}{s(s+2)(s^2+2s+2)} = \frac{10 \cdot 3 \left(\frac{s}{3} + 1\right)}{s \cdot 2 \left(\frac{s}{2} + 1\right) \cdot 2 \left(\frac{s^2}{2} + \frac{2s}{2} + 1\right)} = \frac{7.5(0.33s+1)}{s(0.5s+1)(0.5s^2+s+1)}$$

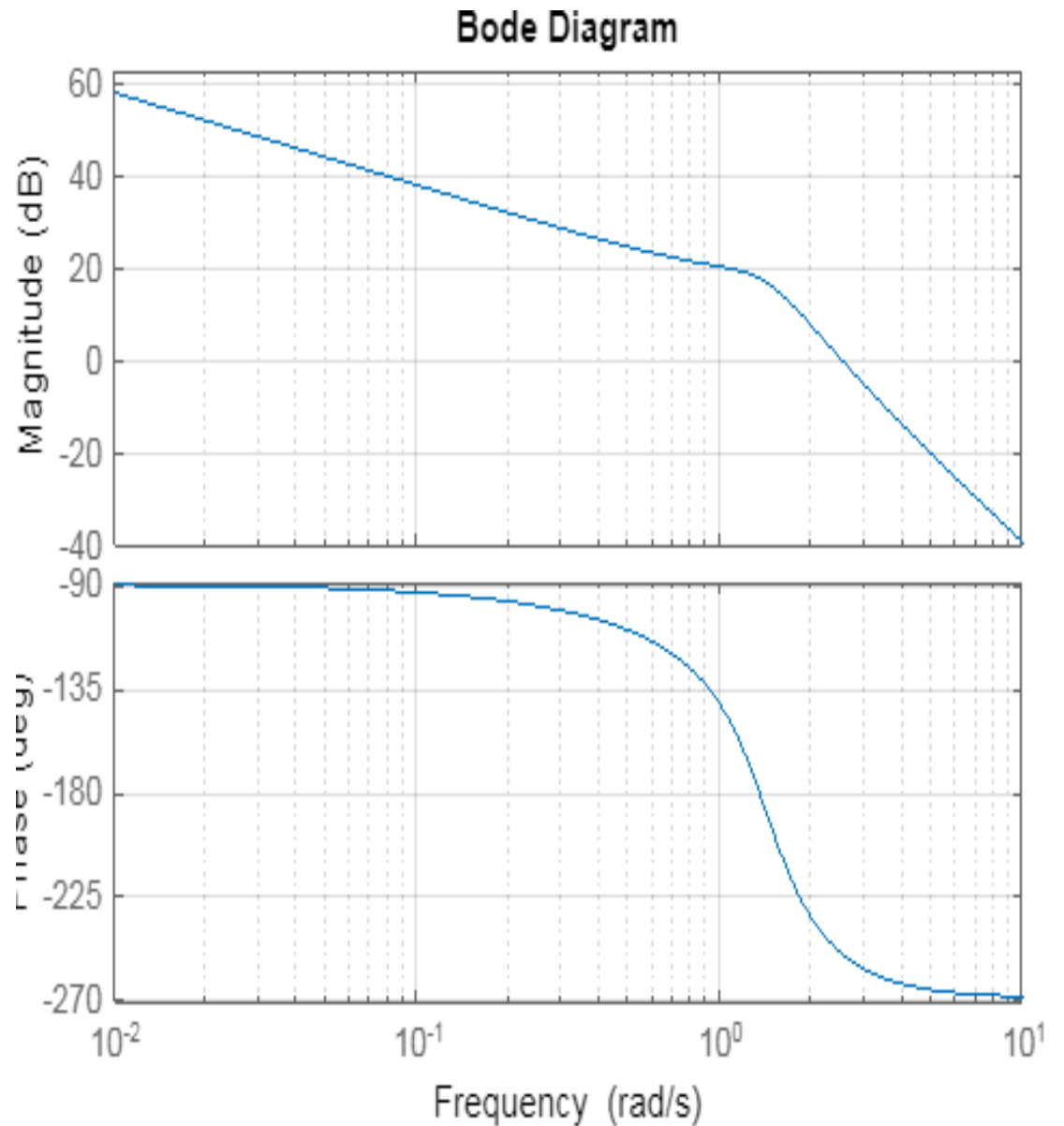
$$S = j\omega$$

$$G(j\omega) = \frac{7.5(0.33j\omega+1)}{j\omega(0.5j\omega+1)(0.5(j\omega)^2+j\omega+1)} = \frac{7.5(0.33j\omega+1)}{j\omega(0.5j\omega+1)(1-0.5\omega^2+j\omega)}$$

$$M_{db} = 20\log 7.5 + 20\log \sqrt{1+(0.33\omega)^2} - 20\log \omega - 20\log \sqrt{1+(0.5\omega)^2} - 20\log \sqrt{\{(1-0.5\omega^2)^2 + \omega^2\}}$$

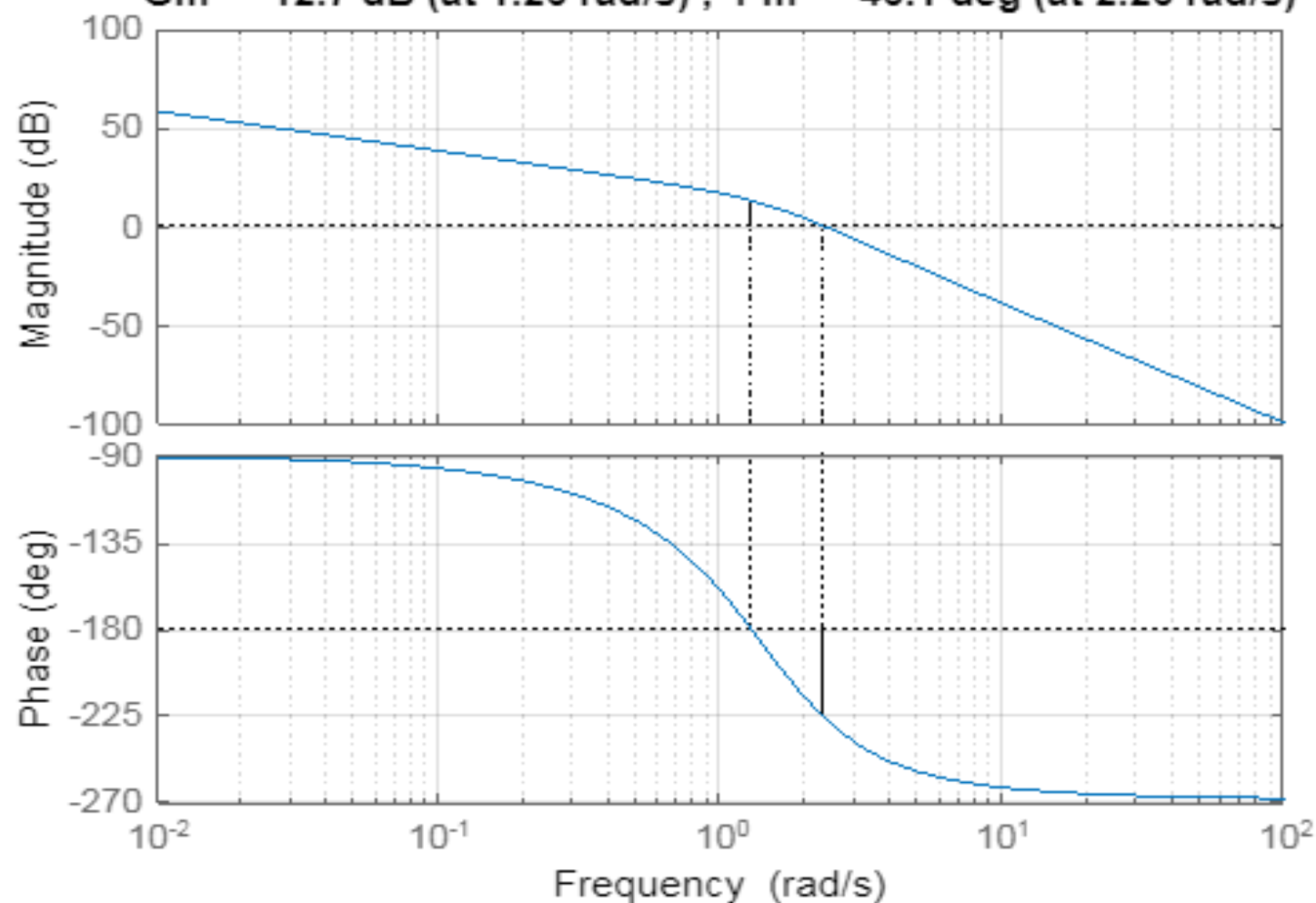
$$\begin{aligned} \emptyset &= \tan^{-1}\left(\frac{0}{7.5}\right) + \tan^{-1}\left(\frac{0.33\omega}{1}\right) - \tan^{-1}\left(\frac{\omega}{0}\right) - \tan^{-1}\left(\frac{0.5\omega}{1}\right) - \tan^{-1}\left(\frac{\omega}{1-0.5\omega^2}\right) \\ &= 0 + \tan^{-1} 0.33\omega - 90 - \tan^{-1} 0.5\omega - \tan^{-1}\left(\frac{\omega}{1-0.5\omega^2}\right) \end{aligned}$$

| ω | M | ϕ |
|----------|-------|-----------------------|
| 0.1 | 63.5 | -96.7 |
| 1 | 23 | -161.7 |
| 2 | 6 | -38(-180) -218 |
| 3 | -5.9 | -61(-180) -241 |
| 10 | -39.7 | -84(-180) -264 |
| 100 | -100 | -89.4(-180) -269.4 |



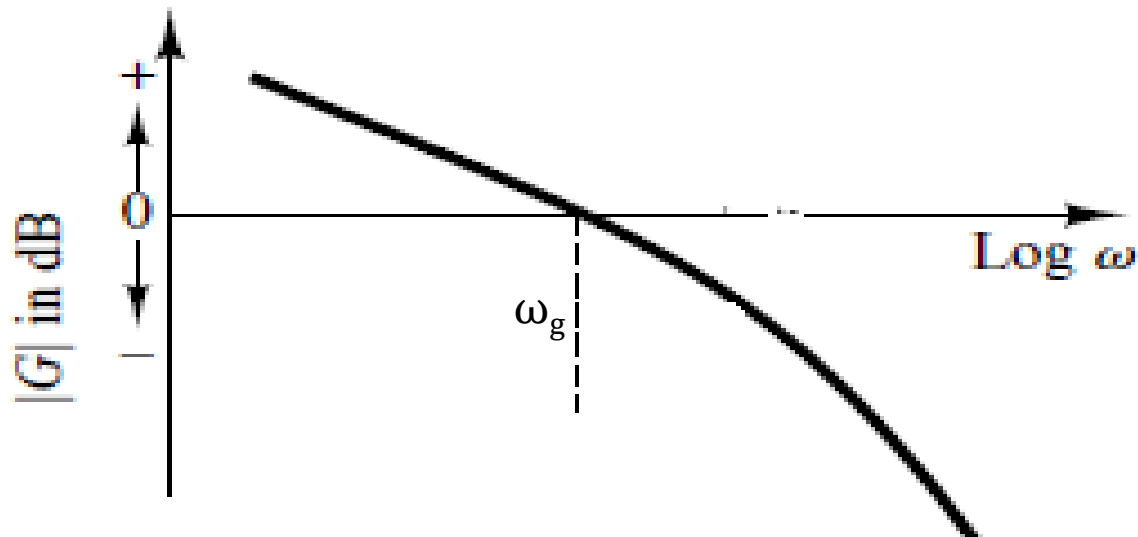
Bode Diagram

Gm = -12.7 dB (at 1.26 rad/s) , Pm = -46.1 deg (at 2.26 rad/s)



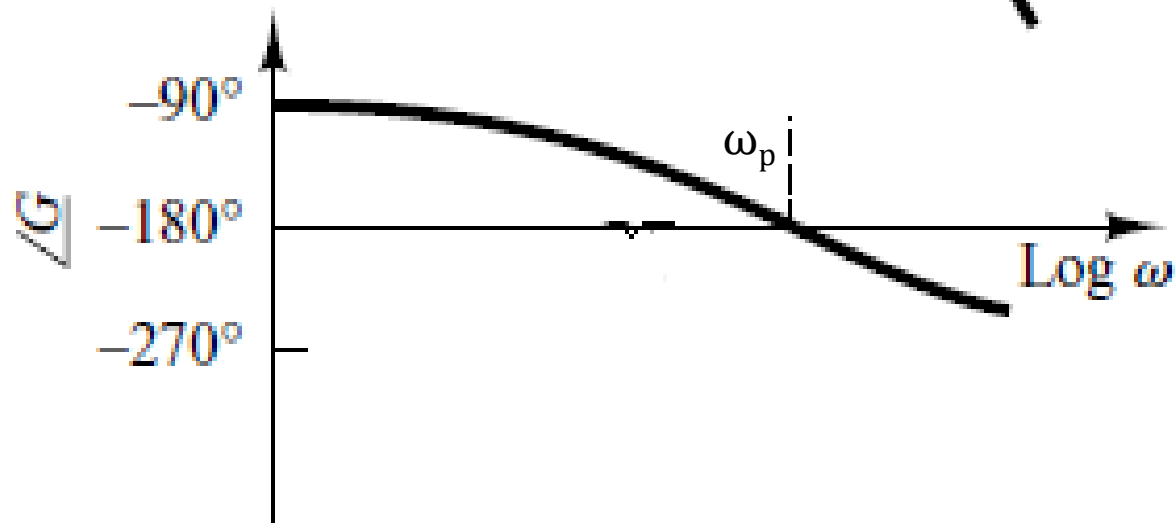
Gain Crossover frequency

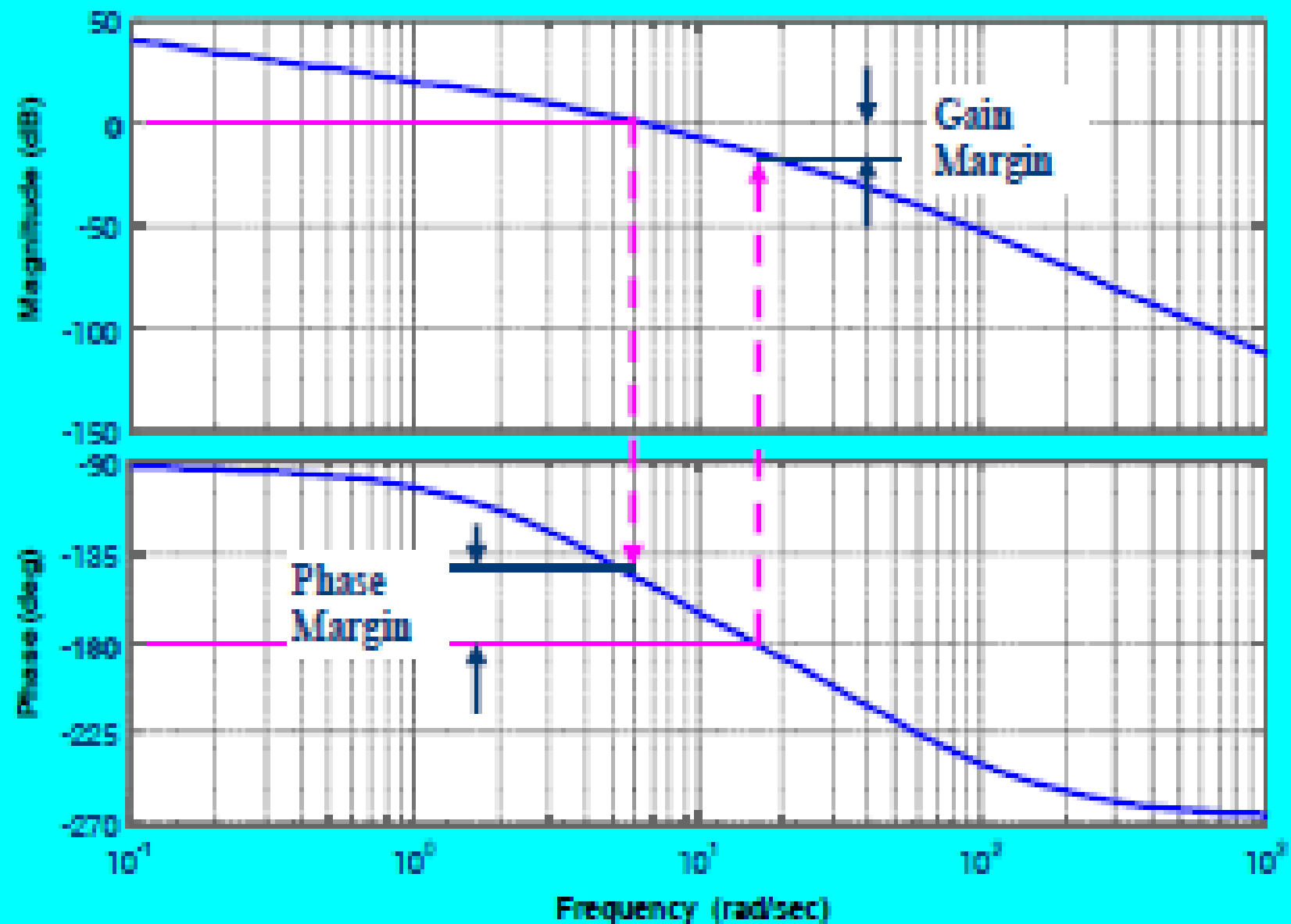
The gain crossover frequency is the frequency at which the magnitude of the open loop transfer function, $|G(j\omega)|$ is unity. In logarithmic scale this corresponds to zero db



Phase Cross over frequency

The phase crossover frequency is the frequency at which the phase of the open loop transfer function, $\angle G(j\omega)$ is -180° .





Gain margin

- The gain margin is the amount of additional gain required at the phase crossover frequency to bring the system to the verge of instability.
- The gain margin is the reciprocal of the magnitude $|G(j\omega)|$ at the phase crossover frequency.

$$GM = \frac{1}{|G(j\omega_p)|}$$

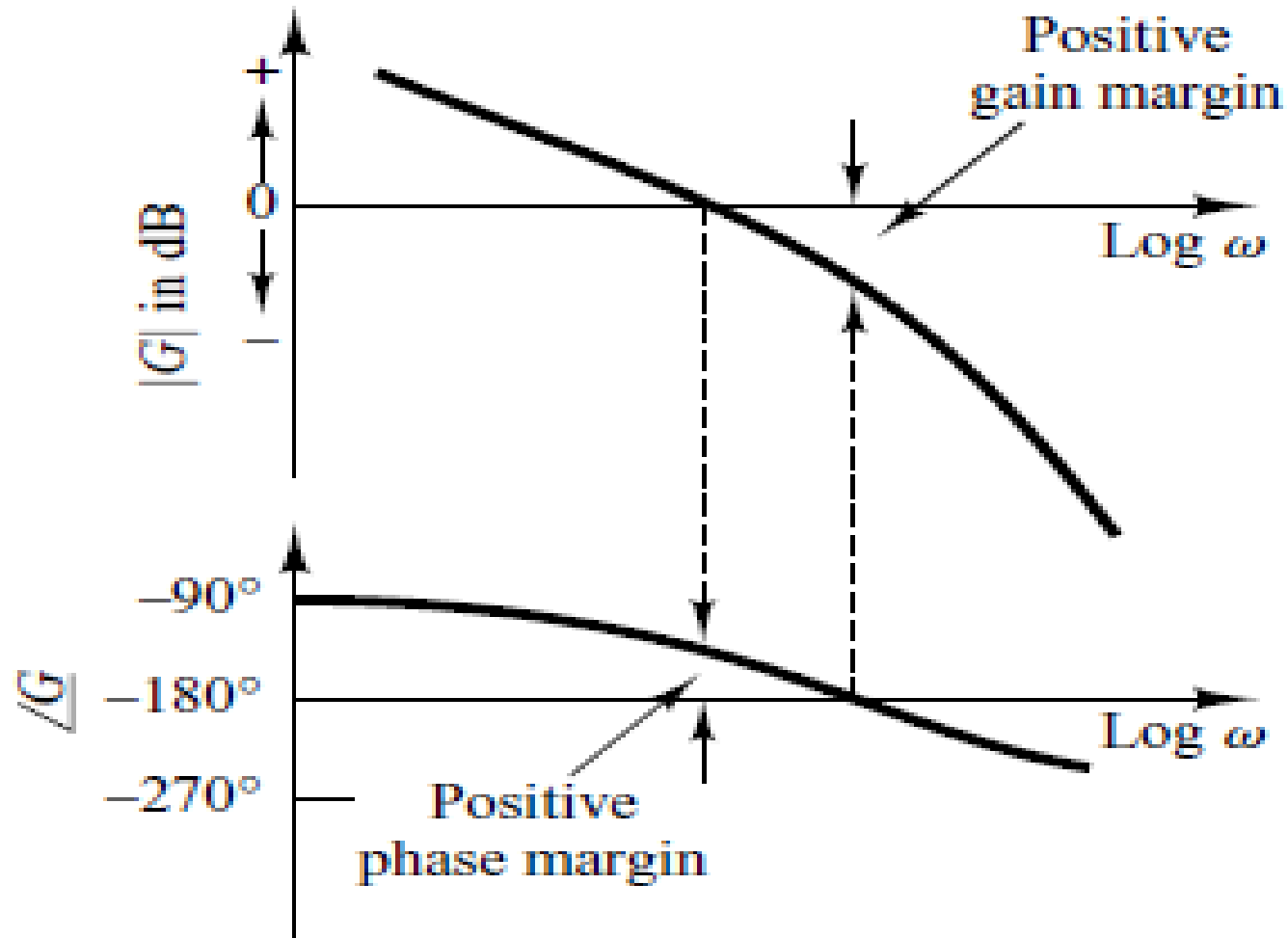
$$\text{In decibels, } GM = -20\log |G(j\omega_p)|$$

Phase margin

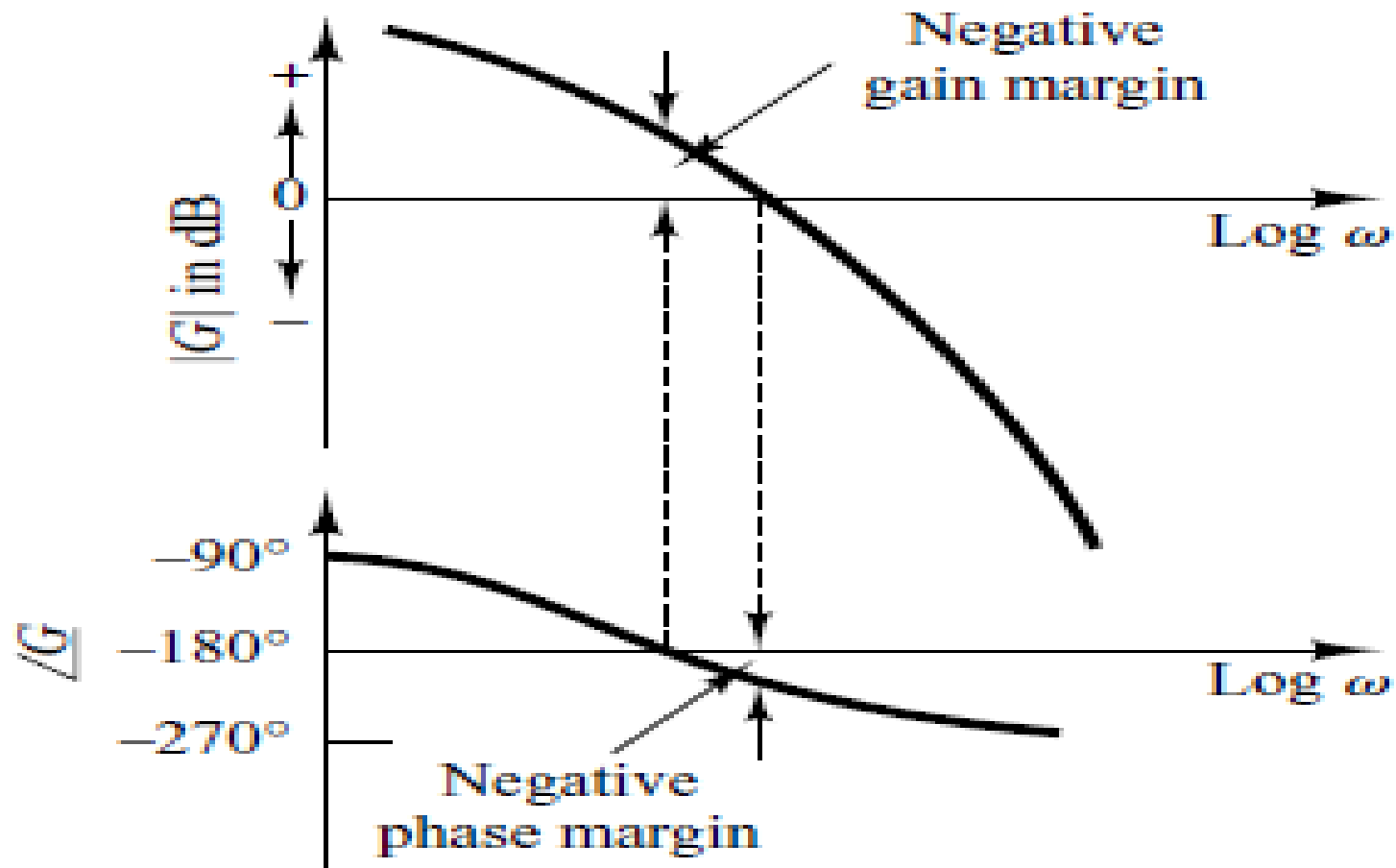
- The phase margin is the amount of additional phase lag required at the gain crossover frequency to bring the system to the verge of instability.
- The phase margin is 180° plus the phase angle of the open-loop transfer function at the gain crossover frequency ω_g

$$PM = 180^\circ + \phi$$

Stable System



Unstable System

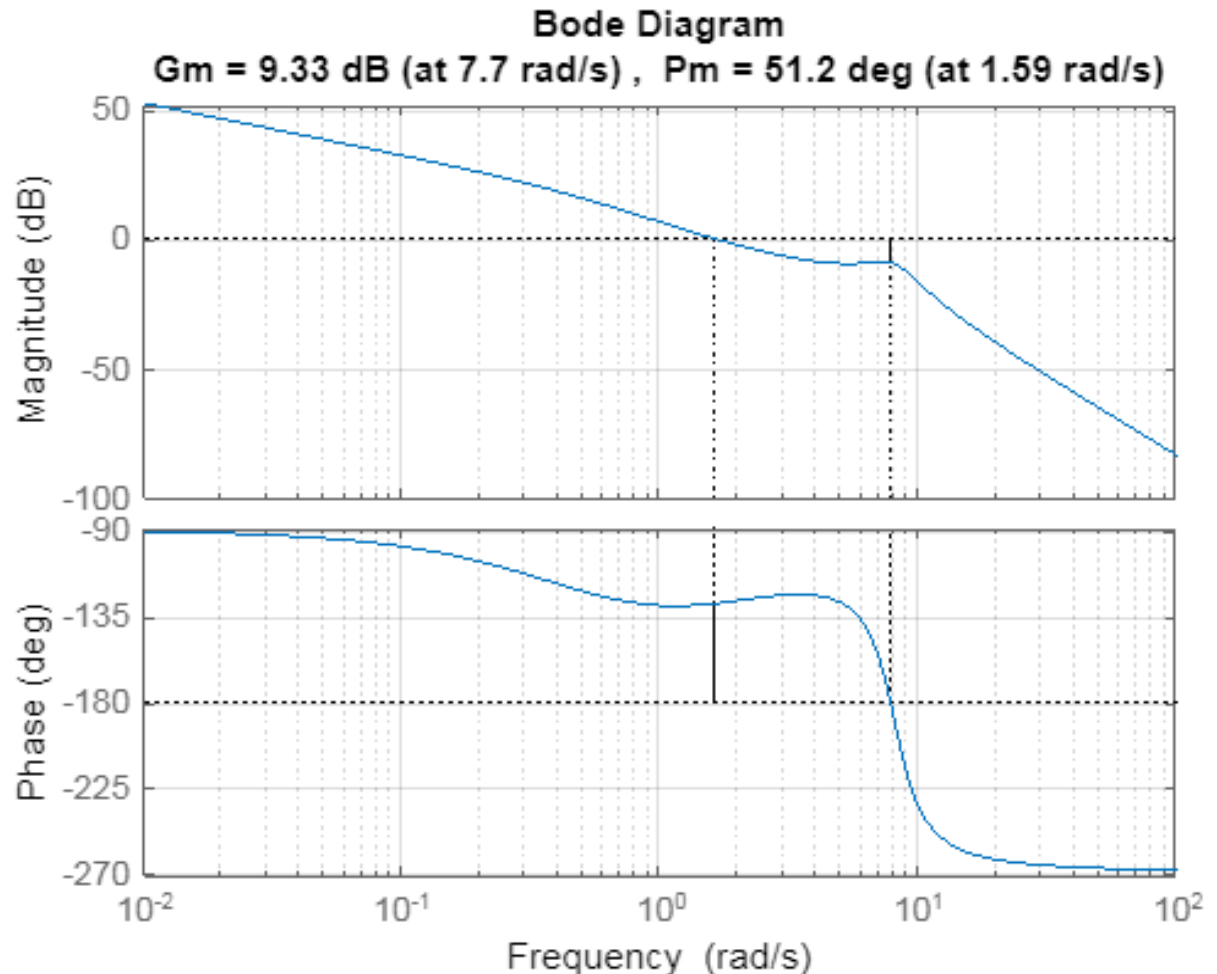


Problem

The open loop transfer function of a unity feedback system is given by

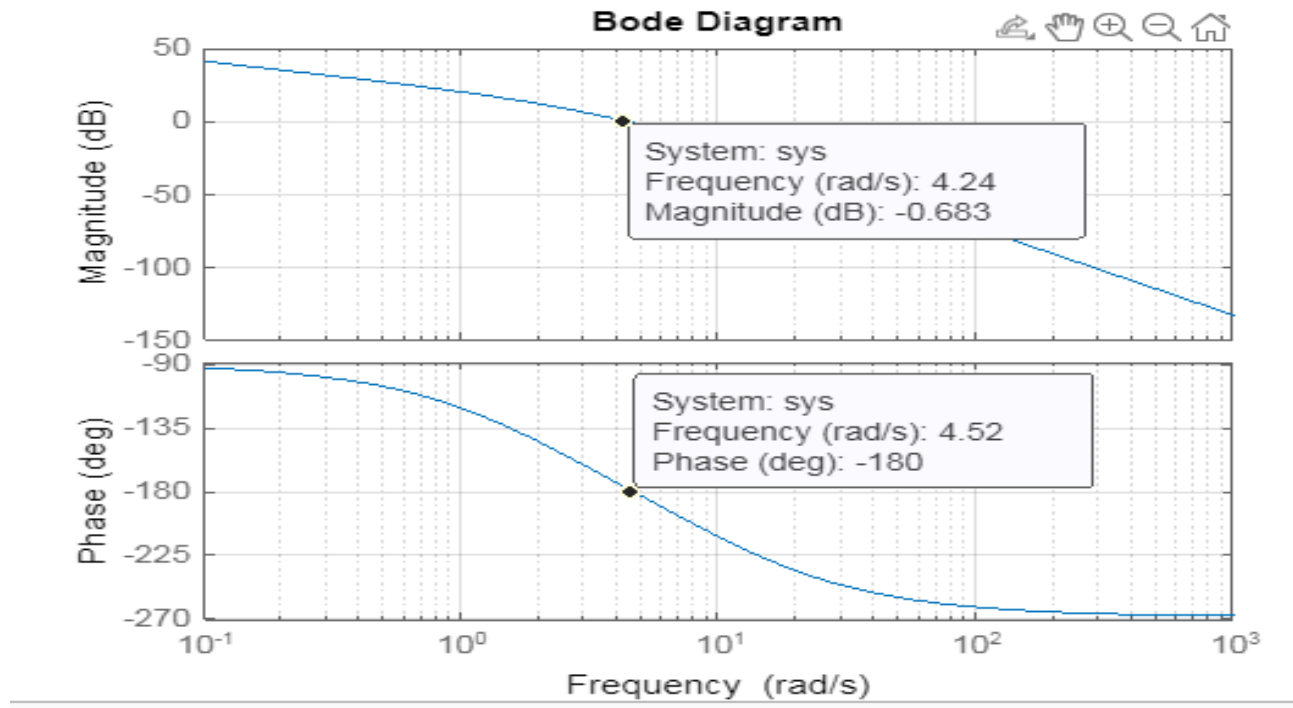
$\frac{64(s+2)}{s(s+0.5)(s^2+3.2s+64)}$. Draw the Bode plot and hence comment on stability.

```
num=[64 128];  
den=[1 3.7 65.6 32 0];  
bode(num,den)  
margin(num,den)
```

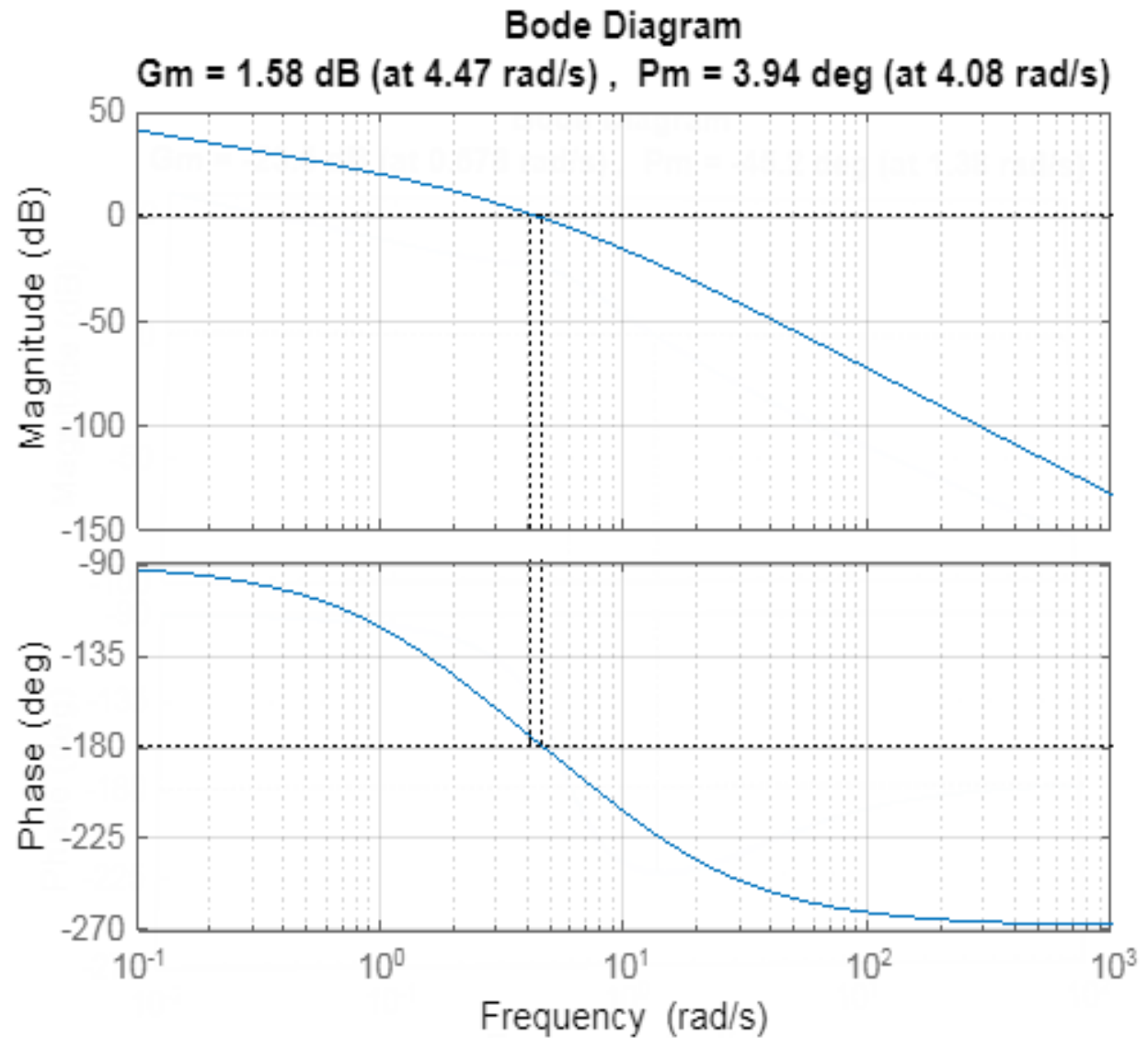


Problem

Draw the Bode plot for the transfer function $G(s) = \frac{10}{s(1+0.5s)(1+0.1s)}$ and hence comment on stability.

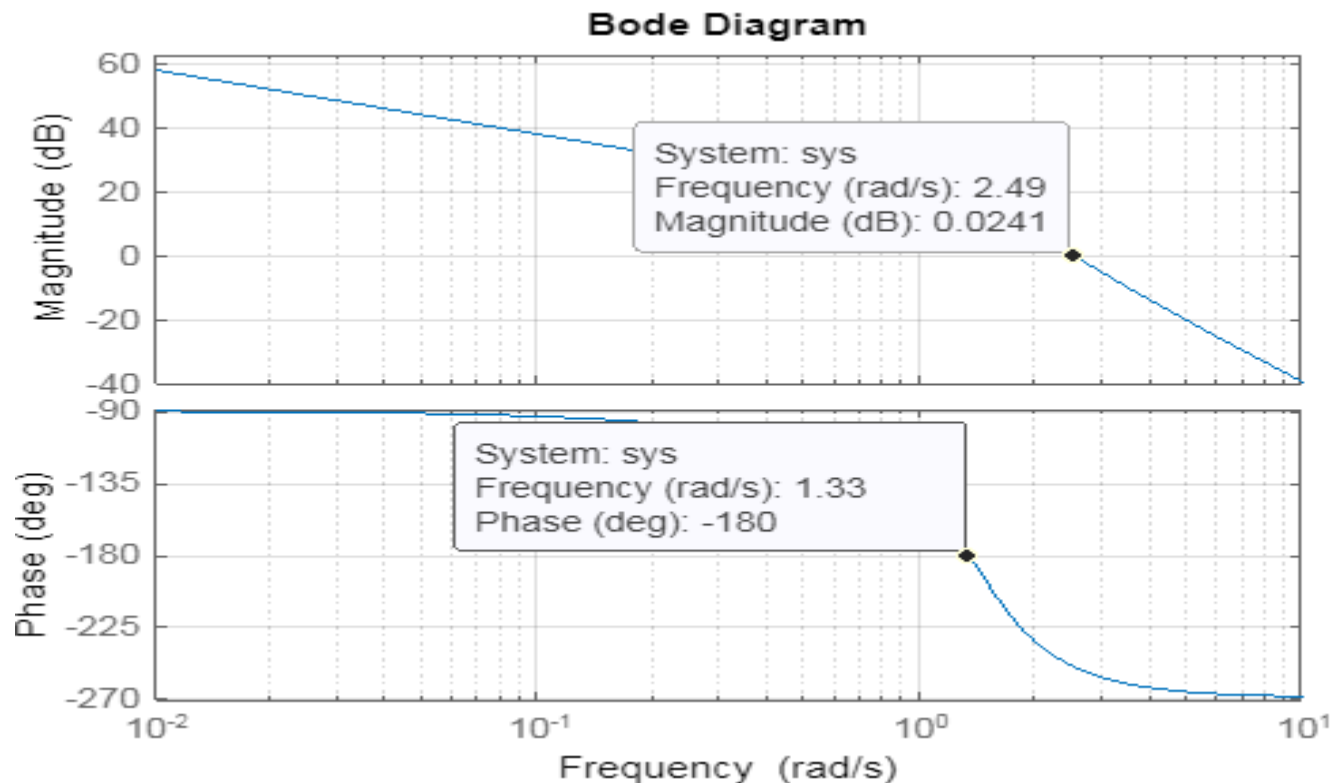


```
num=[10];  
den=[0.05 0.6 1 0];  
bode(num,den)  
margin(num,den)  
grid on
```



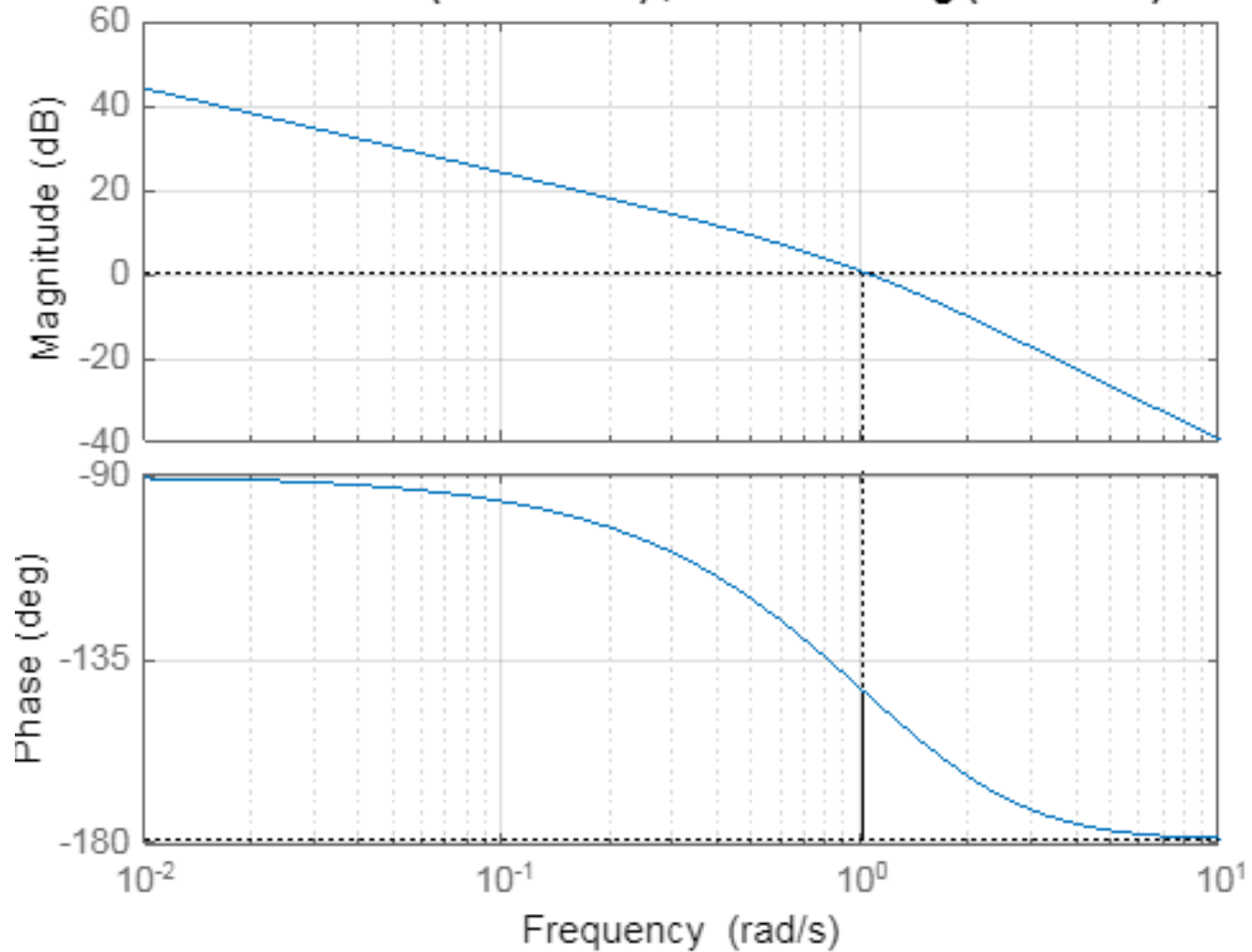
Problem

Draw the Bode plot for the transfer function $G(s) = \frac{(s+3)}{s(s+1)(s+2)}$ and hence comment on stability.



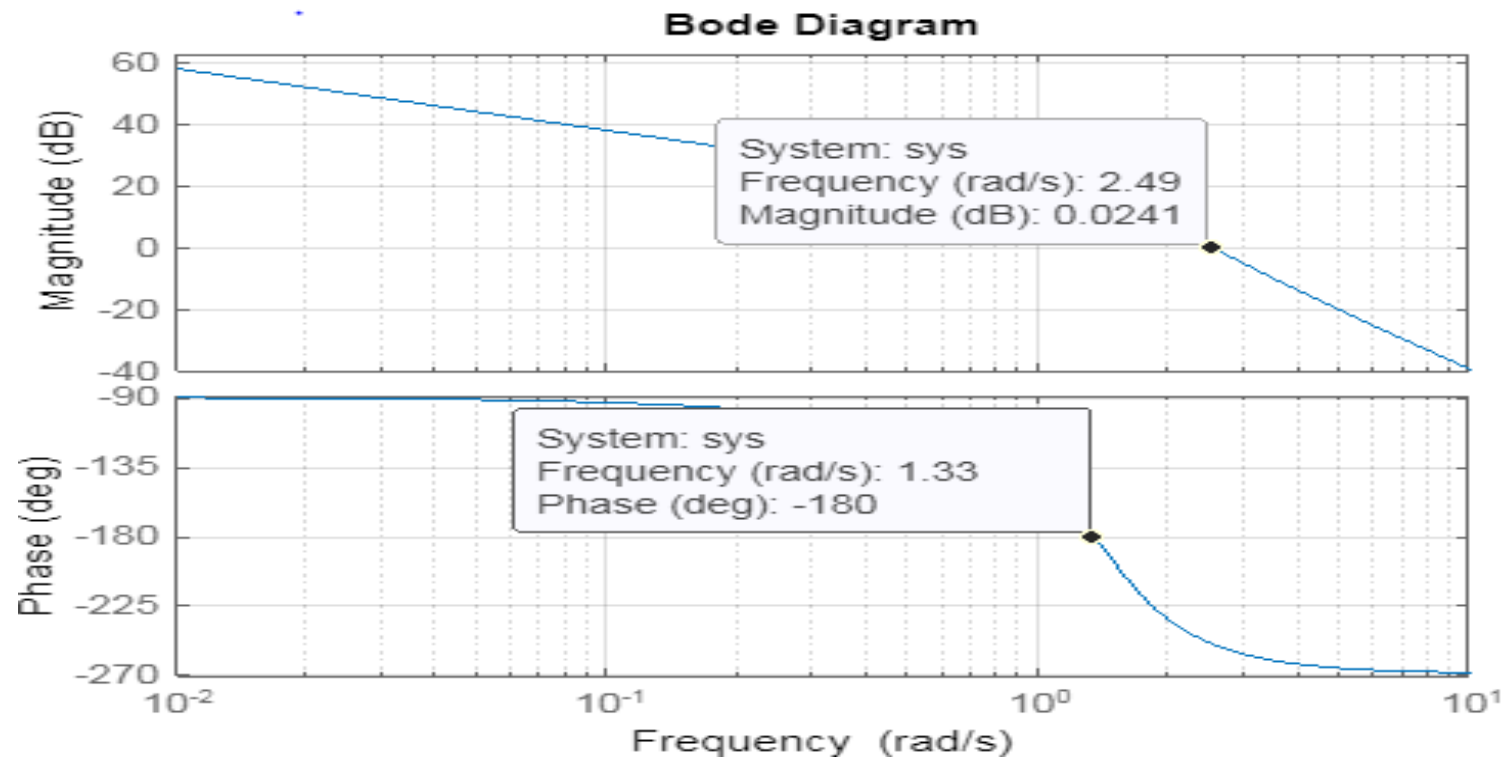
Bode Diagram

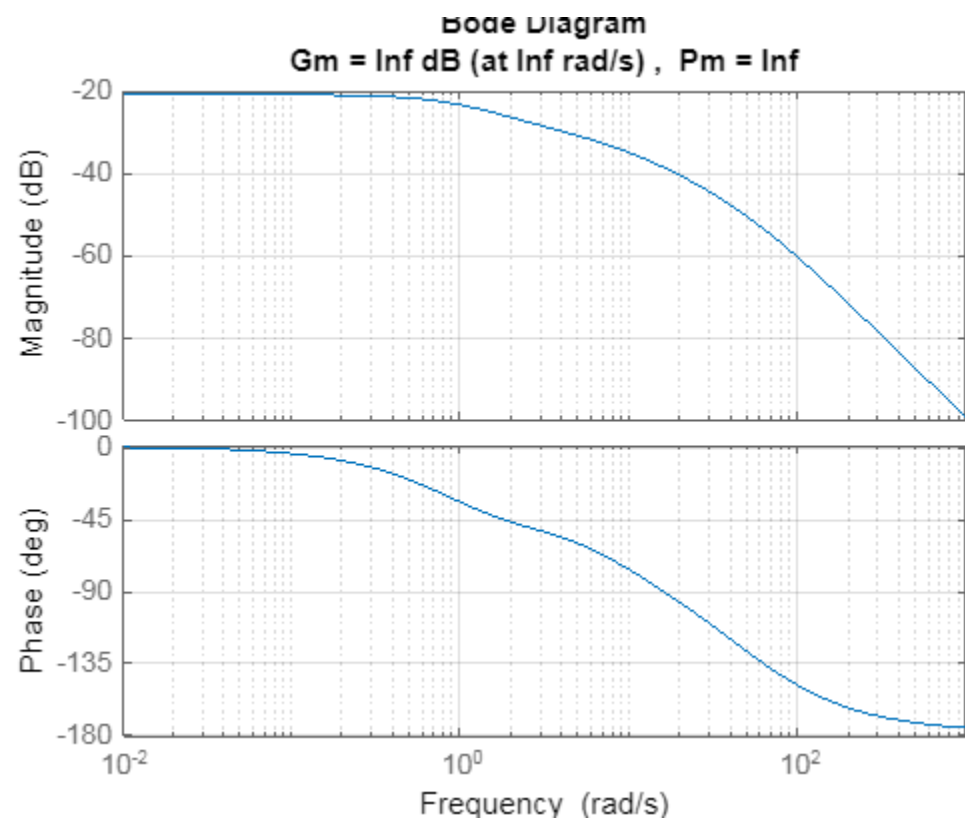
$G_m = \infty \text{ dB (at } \infty \text{ rad/s) , } P_m = 36.9 \text{ deg (at } 1 \text{ rad/s)}$



Problem 5

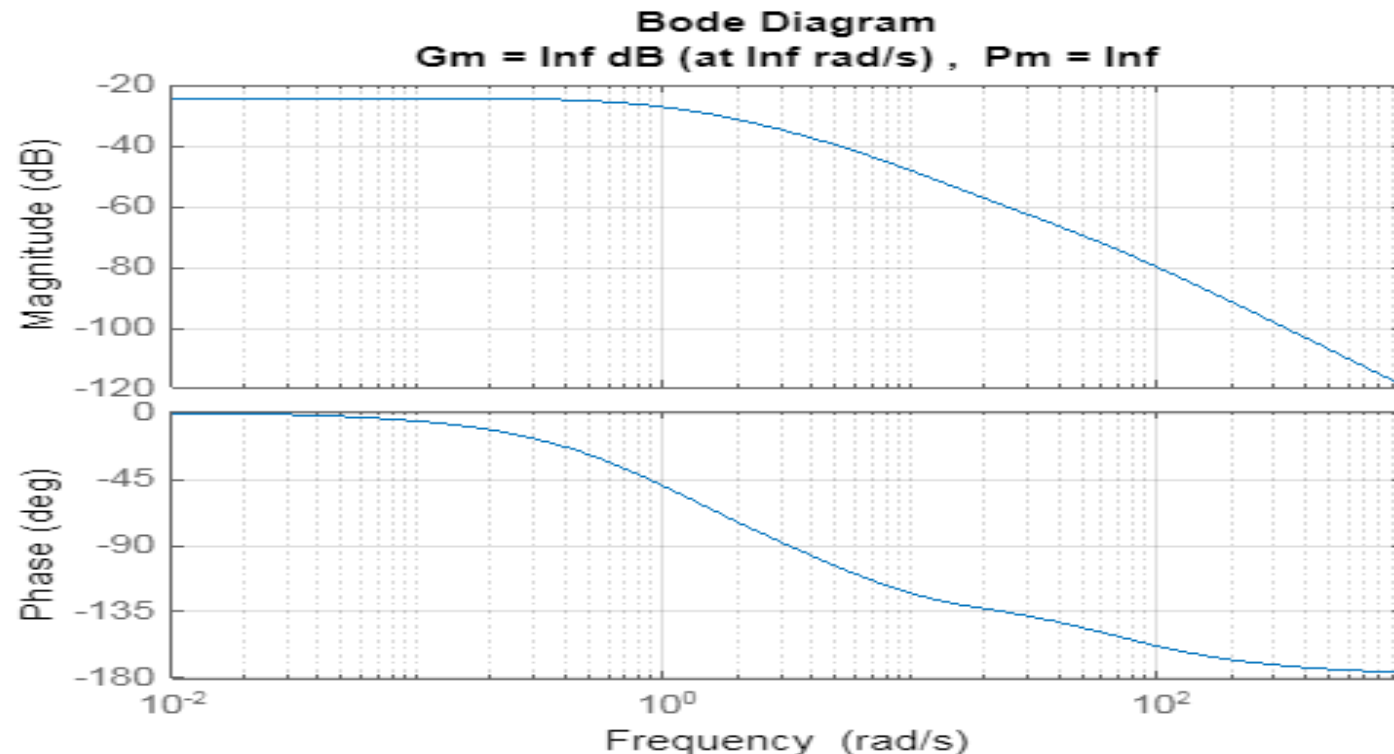
Comment on stability for the following transfer function using Bode plot $\frac{10(s+3)}{(s+1)(s+7)(s+50)}$





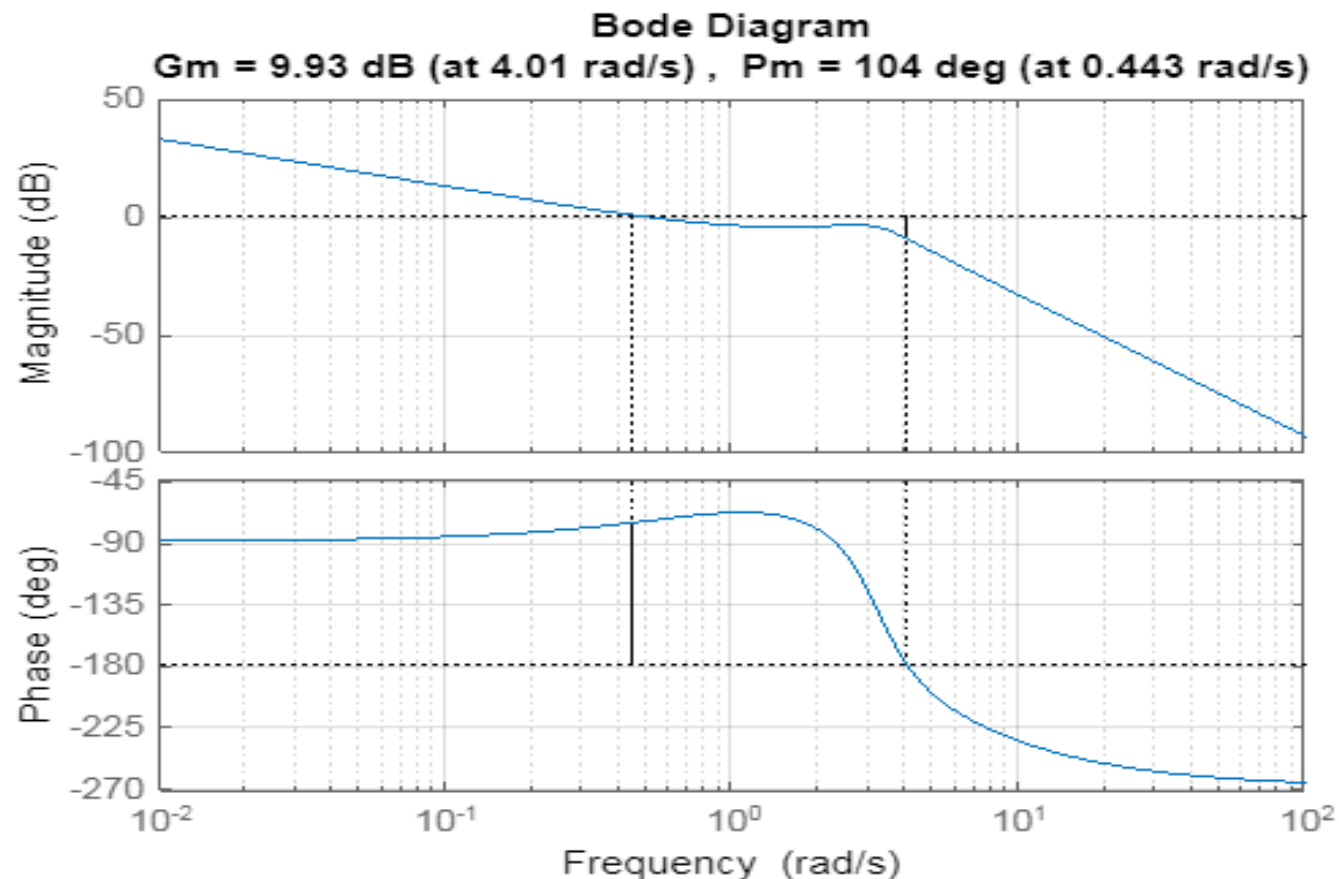
Problem 6

Draw the Bode log-magnitude and phase plots
for the system $\frac{(s+20)}{(s+1)(s+7)(s+50)}$



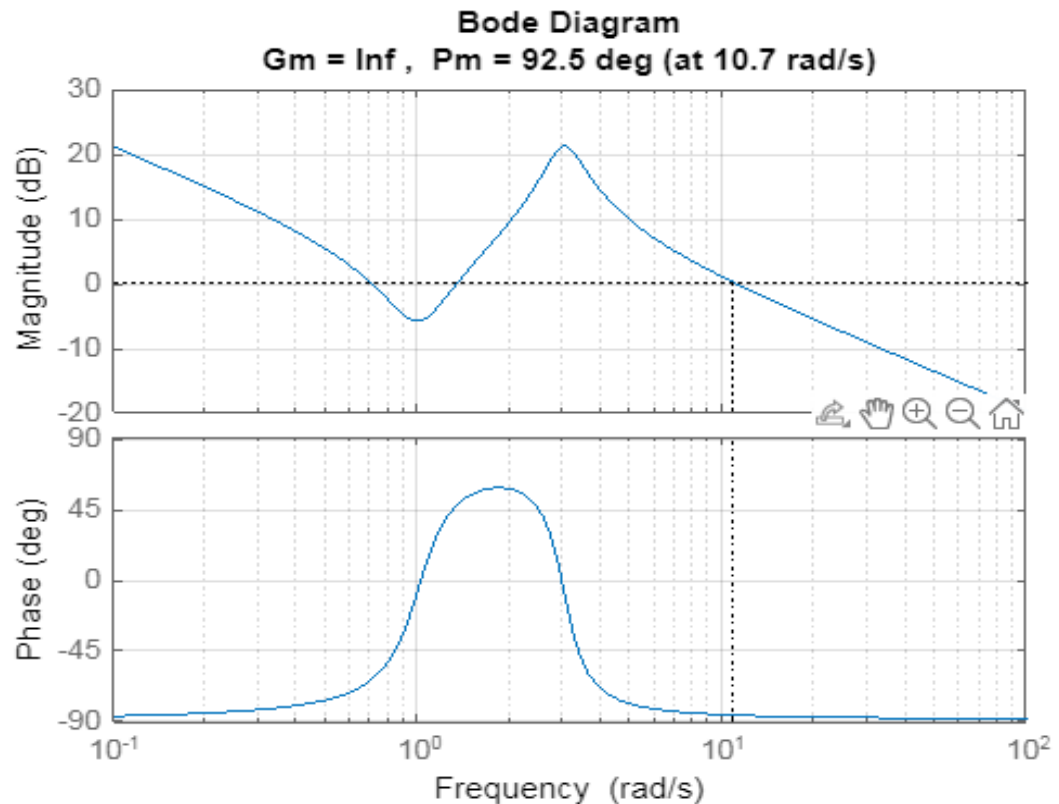
Problem 7

Draw the Bode log-magnitude and phase plots for the system $\frac{20(s+1)}{s(s+5)(s^2+2s+10)}$



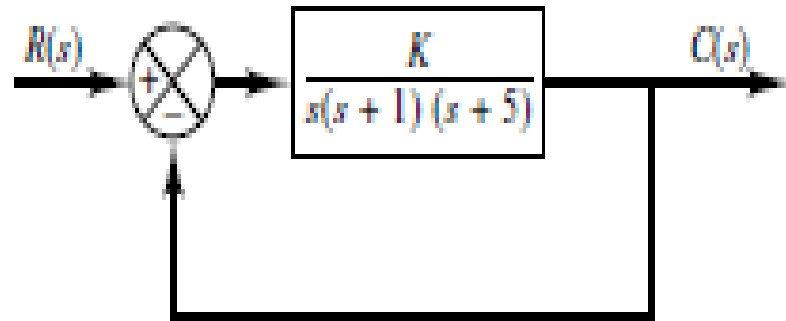
Problem 8

The open loop transfer function of a unity feedback system is given by $\frac{10(s^2+0.4s+1)}{s(s^2+0.8s+9)}$. Draw the Bode plot

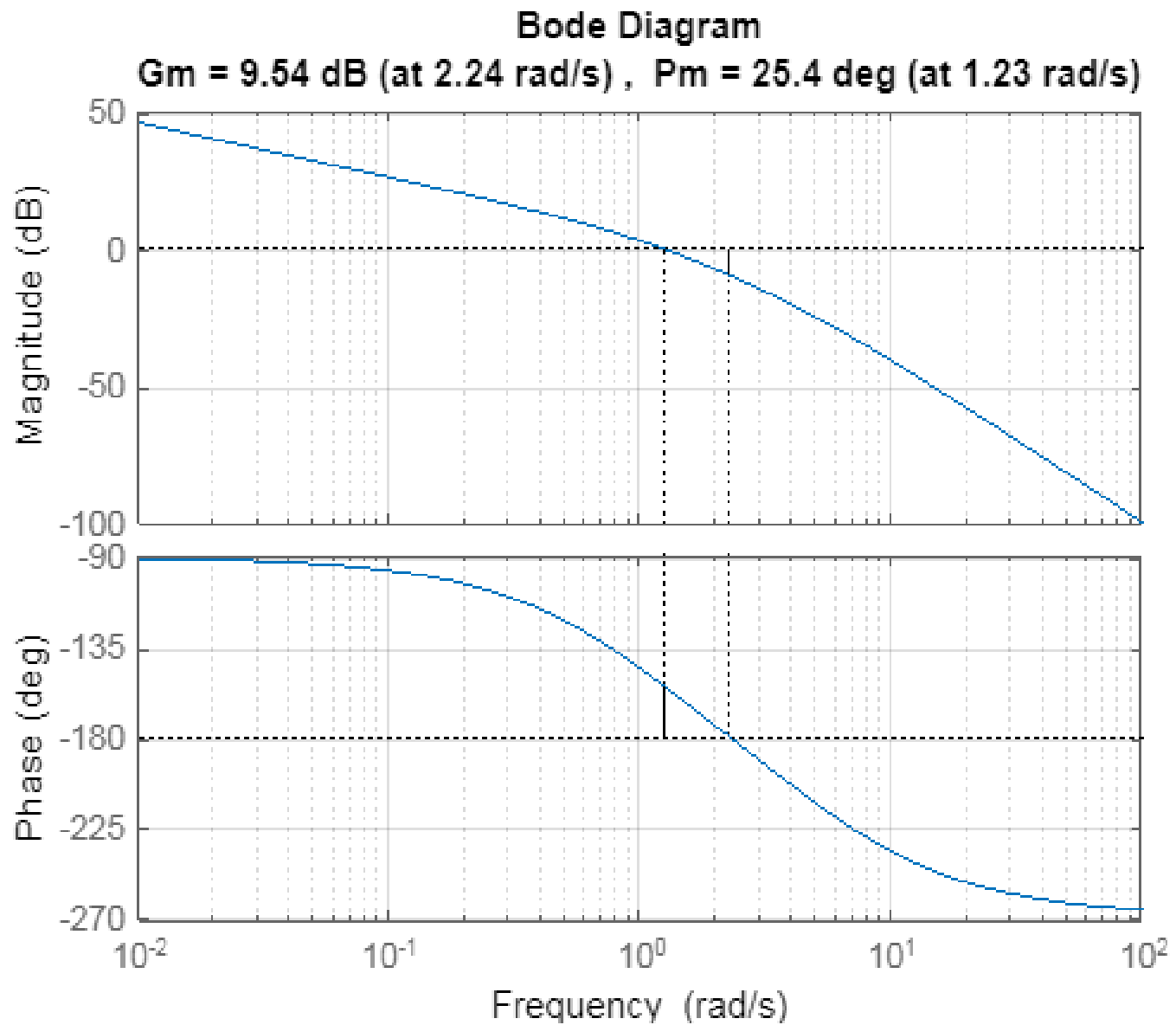


Problem 9

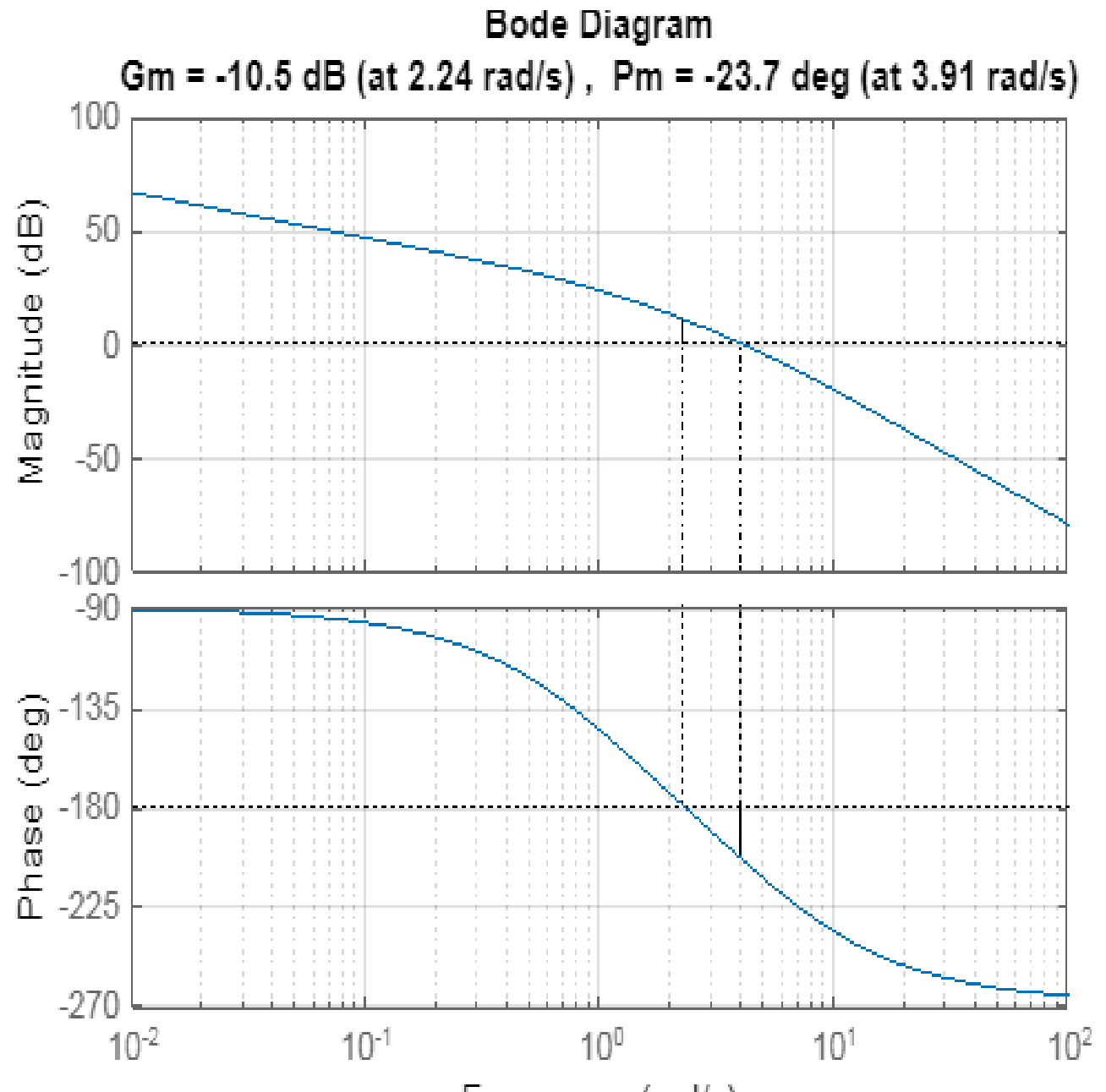
Obtain the phase and gain margins of the system shown in Figure for the two cases where $K=10$ and $K=100$



$K=10$



K=100



Problem

Draw the approximate (Asymptotic) Bode plot for the open loop transfer function $G(s) = \frac{(s+3)}{s(s+1)(s+2)}$

$$G(s) = \frac{(s+3)}{s(s+1)(s+2)} = \frac{3(\frac{s}{3}+1)}{s(s+1)2(\frac{s}{2}+1)} = \frac{1.5(0.33s+1)}{s(s+1)(0.5s+1)}$$

$$S=j\omega$$

$$G(j\omega) = \frac{1.5(0.33j\omega+1)}{j\omega(j\omega+1)(0.5j\omega+1)}$$

Magnitude

$$|G(j\omega)| = \frac{|1.5| * |(0.33j\omega + 1)|}{|j\omega| * |(j\omega + 1)| * |(0.5j\omega + 1)|}$$

$$M_{db} = \frac{20\log |1.5| + 20\log |(0.33j\omega + 1)|}{20\log |j\omega| + 20\log |(j\omega + 1)| + 20\log |(0.5j\omega + 1)|}$$

$$= 20\log |1.5| + 20\log |(0.33j\omega + 1)| - 20\log |j\omega| \\ - 20\log |(j\omega + 1)| - 20\log |(0.5j\omega + 1)|$$

$$= M_1 + M_2 + M_3 + M_4 + M_5$$

$$M_1 = 20\log 1.5 = 3.5 \text{ dB}$$

constant

$$M_2 = 20\log |(0.33j\omega + 1)|$$

zero at -3 rad/sec

Corner frequency 3 rad/sec

magnitude is 0 dB upto CF

above CF, slope = 20 dB/dec

$$M_3 = -20 \log |j\omega|$$

Pole at origin

Slope = -20dB/dec

$$M_4 = -20 \log |j\omega + 1|$$

Pole at -1

Corner frequency = 1 rad/sec

Magnitude = 0 dB upto 1 rad/sec,

Above CF, slope = -20 dB/dec

$$M_5 = -20 \log |0.5j\omega + 1|$$

Pole at -2

Corner frequency = 2 rad/sec

Magnitude = 0 dB upto 2 rad/sec,

Above CF, slope = -20 dB/dec

Phase

$$\begin{aligned}\emptyset &= \angle 1.5 + \angle(1+0.33j\omega) - \angle j\omega - \angle(1+j\omega) - \angle(1+0.5j\omega) \\ &= 0 + \angle(1+0.33j\omega) - 90 - \angle(1+j\omega) - \angle(1+0.5j\omega) \\ &= \emptyset_1 + \emptyset_2 + \emptyset_3 + \emptyset_4 + \emptyset_5\end{aligned}$$

$$\emptyset_1 = \angle 1.5 = 0 \text{ degrees} \quad (\text{constant})$$

$$\emptyset_2 = \angle(1+0.33j\omega) \quad (\text{zero at 3rad/sec})$$

Corner frequency = 3 rad/sec

At CF, $\emptyset = 45$ degrees, above C.F = 90 degrees

$$\emptyset_3 = -\angle j\omega \quad (\text{pole at origin})$$

$$= -90 \text{ degrees}$$

$$\emptyset_4 = -\angle(1+j\omega) \quad (\text{pole at 1 rad/sec})$$

$$\text{C.F} = 1 \text{ rad/sec}$$

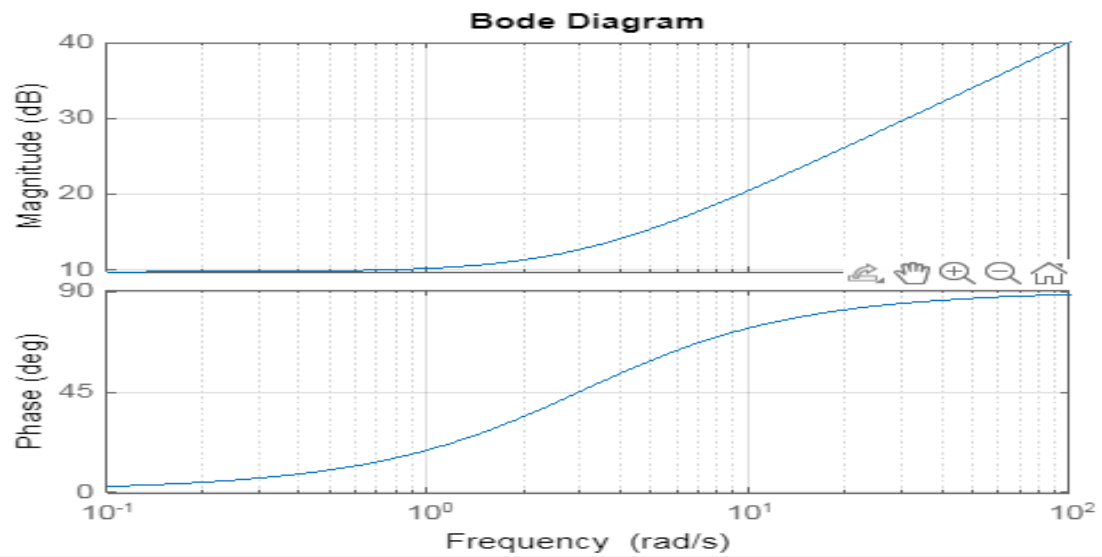
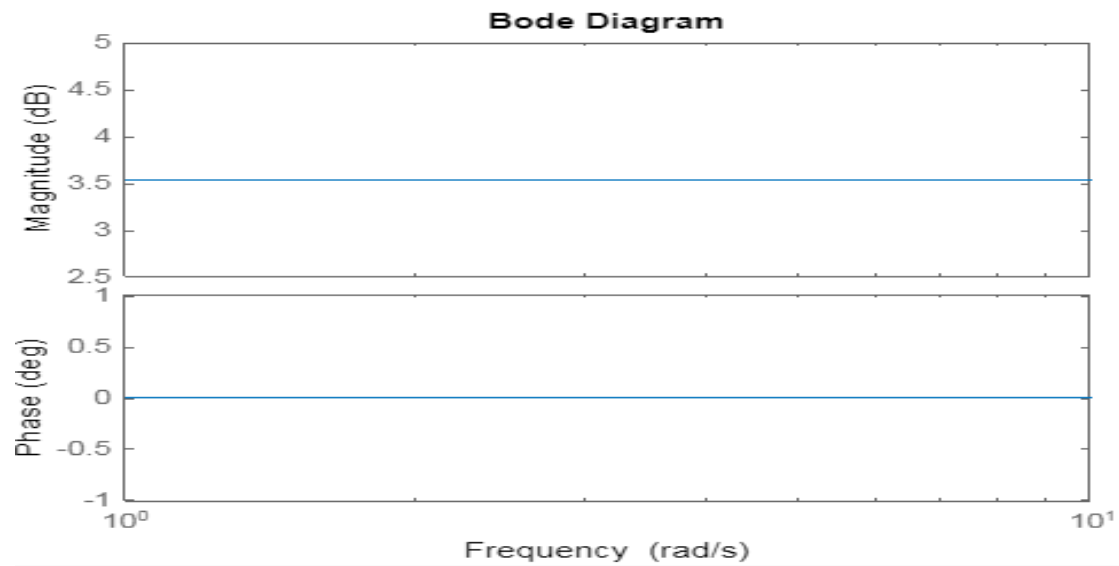
At CF, $\emptyset = -45$ degrees, above C.F = -90 degrees

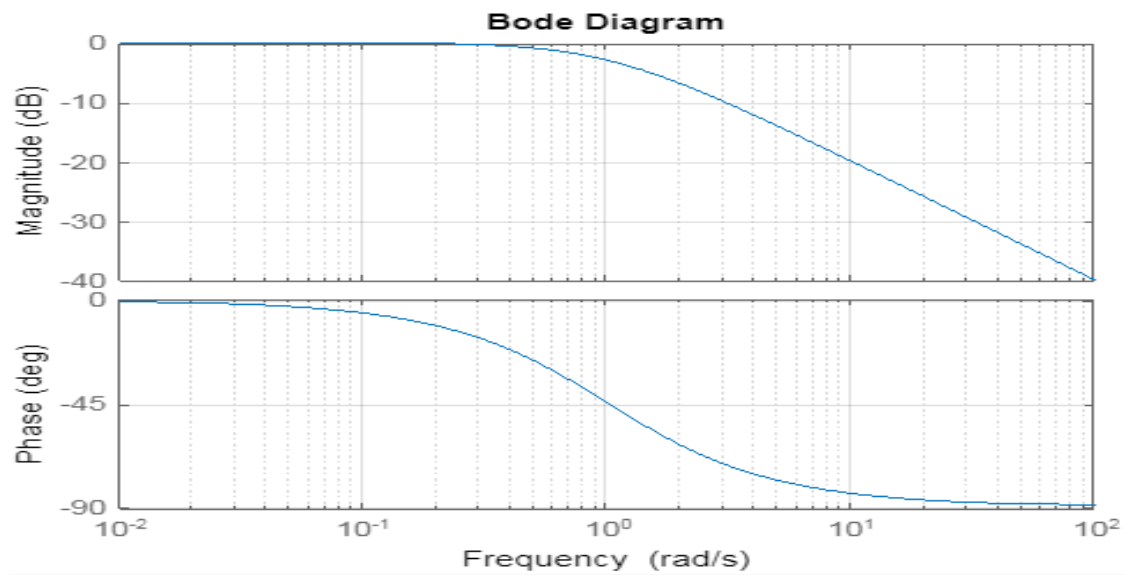
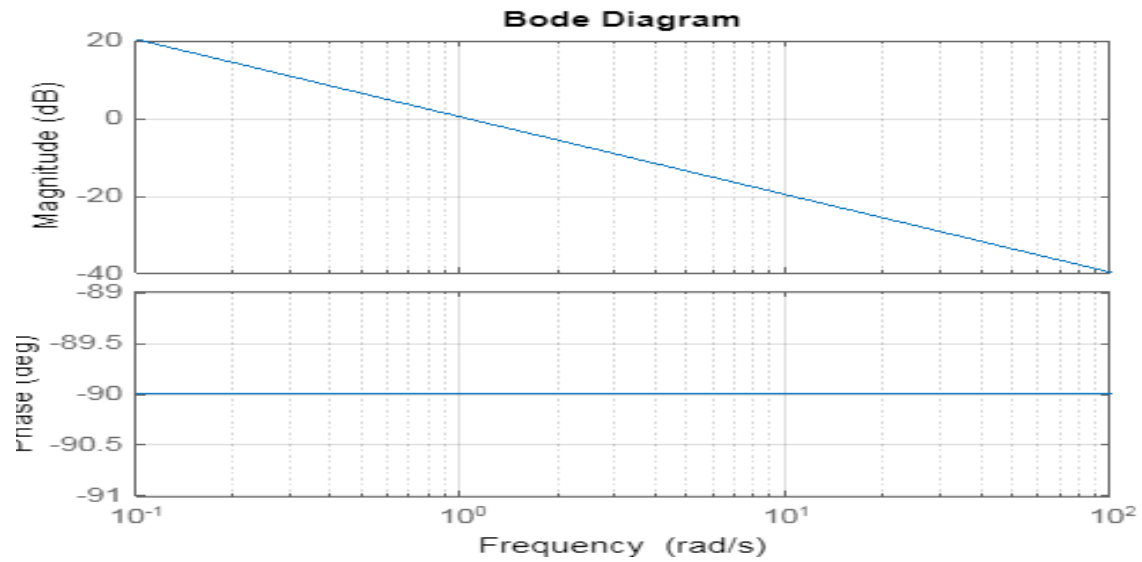
$$\emptyset_5 = -\angle(1+0.5j\omega) \quad (\text{pole at 2 rad/sec})$$

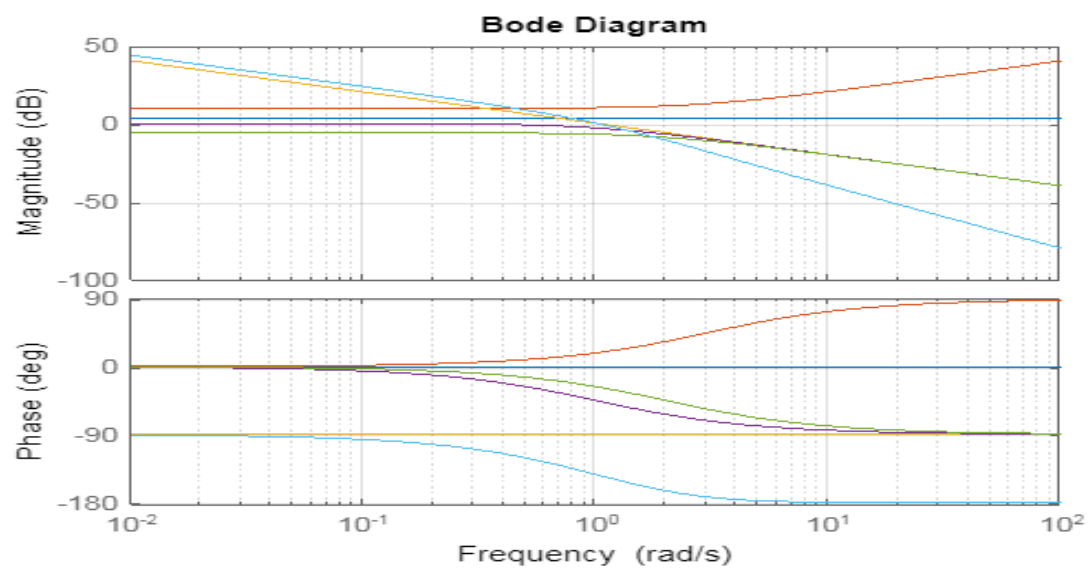
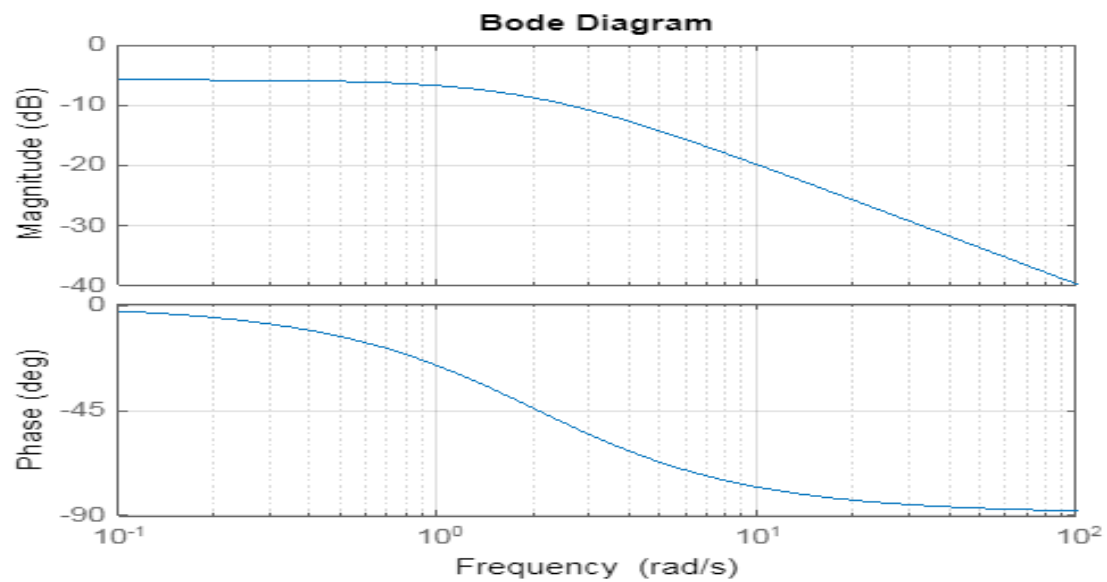
$$\text{C.F} = 2 \text{ rad/sec}$$

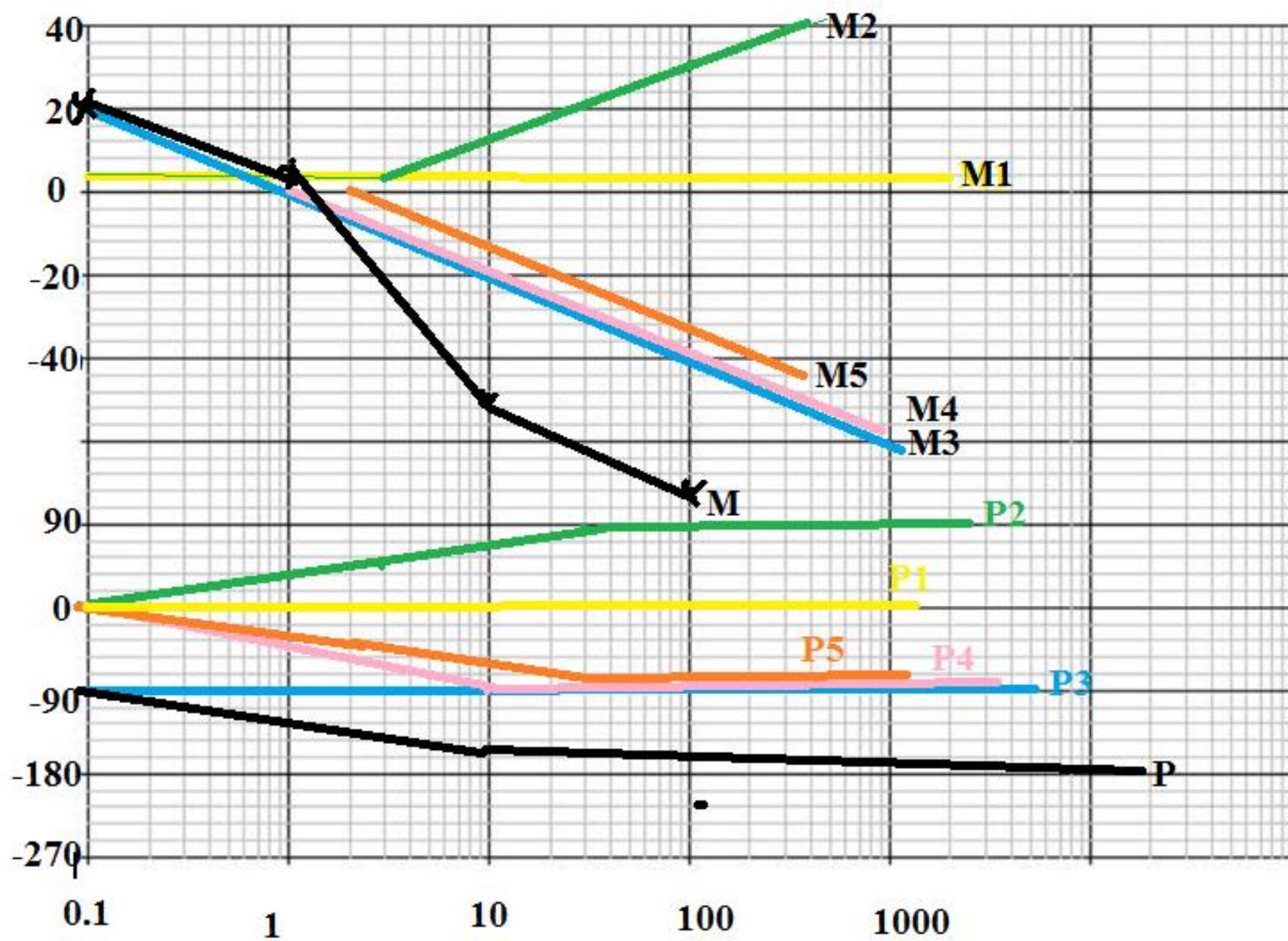
At CF, $\emptyset = -45$ degrees, above C.F = -90 degrees

| Factor | Corner Frequency (rad/sec) | Magnitude Characteristics | Phase Characteristics |
|---|----------------------------|---|---|
| 1.5 | - | Straight line of 3.5 db | zero |
| $\left(\frac{1}{j\omega}\right)$ | - | Straight line of constant slope = -20db/decade passing through zero db at $\omega = 1$ | Constant = -90° |
| $\left(\frac{1}{1 + j\omega}\right)$ | $\omega_1 = 1$ | Straight line of 0 db for $\omega < \omega_1$, straight line of slope = -20db/decade for $\omega > \omega_1$ | Varies from 0 to -90°, at $\omega_1 = -45^\circ$ |
| $\left(\frac{1}{1 + j0.5\omega}\right)$ | $\omega_2 = 2$ | Straight line of 0 db for $\omega < \omega_2$, straight line of slope = -20db/decade for $\omega > \omega_2$ | Varies from 0 to -90°, at $\omega_2 = -45^\circ$ |
| $(1 + j0.33\omega)$ | $\omega_3 = 3$ | Straight line of 0 db for $\omega < \omega_3$, straight line of slope = 20db/decade for $\omega > \omega_3$ | Varies from 0 to 90°, at $\omega_3 = 90^\circ$ |

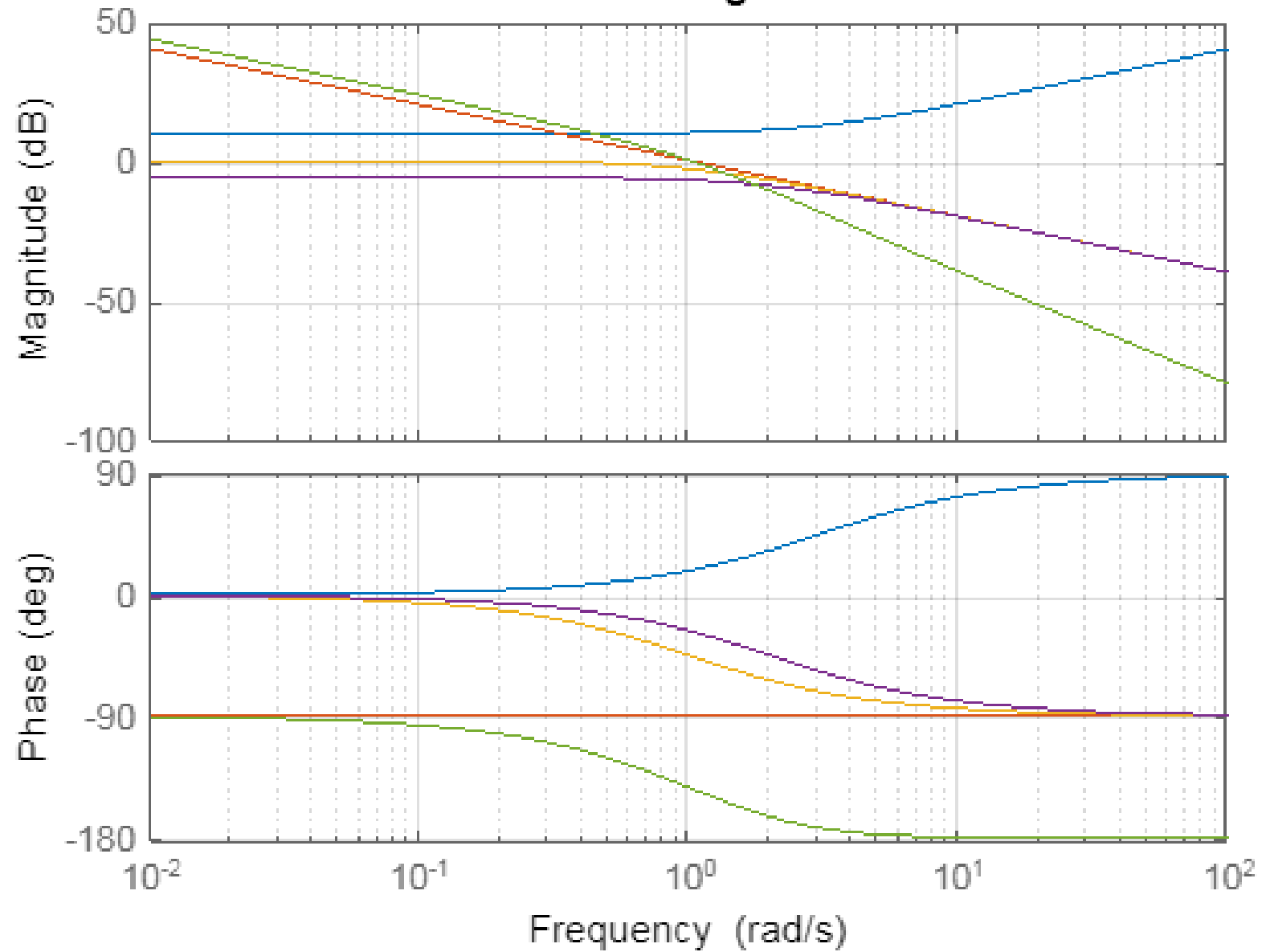






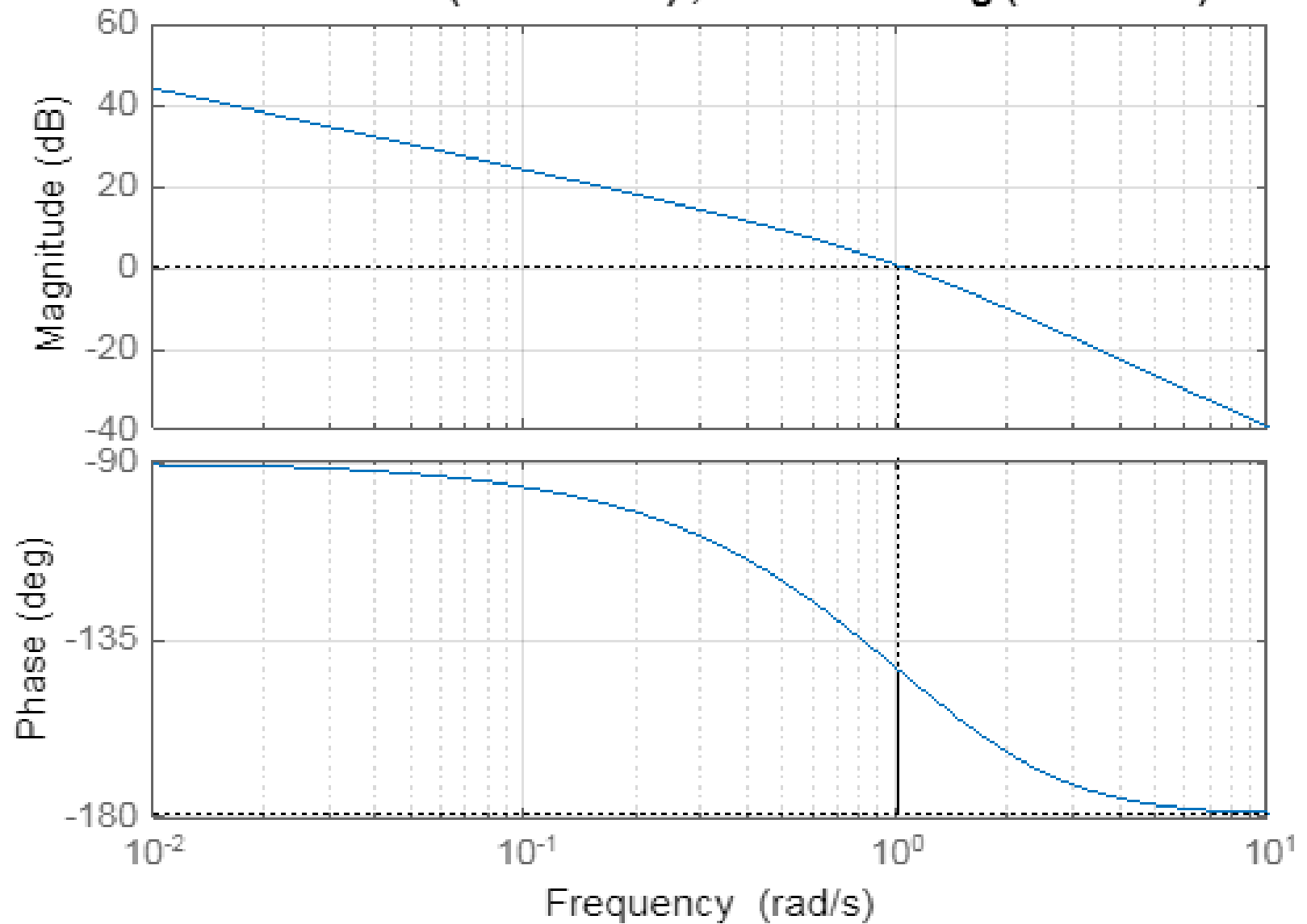


Bode Diagram



Bode Diagram

$G_m = \infty \text{ dB (at } \infty \text{ rad/s)}$, $P_m = 36.9 \text{ deg (at } 1 \text{ rad/s)}$



Problem

The open loop transfer function of a unity feedback system is given by $\frac{64(s+2)}{s(s+0.5)(s^2+3.2s+64)}$. Draw the approximate Bode plot

$$\begin{aligned} G(s) &= \frac{64(s+2)}{s(s+0.5)(s^2+3.2s+64)} = \frac{64 \cdot 2 \left(\frac{s}{2} + 1\right)}{s \cdot 0.5 \left(\frac{s}{0.5} + 1\right) \cdot 64 \cdot \left(\frac{s^2}{64} + \frac{3.2s}{64} + 1\right)} \\ &= \frac{4(0.5s+1)}{s(2s+1)\left(\frac{s^2}{64} + \frac{3.2}{8 \cdot 8}s + 1\right)} = \frac{4(0.5s+1)}{s(2s+1)\left(\frac{s^2}{64} + \frac{0.4}{8}s + 1\right)} \end{aligned}$$

$$S = j\omega$$

$$G(j\omega) = \frac{4(0.5j\omega+1)}{j\omega(2j\omega+1)\left(\left(\frac{j\omega}{8}\right)^2 + j0.4\left(\frac{\omega}{8}\right) + 1\right)} = \frac{4(0.5j\omega+1)}{j\omega(2j\omega+1)\left(-\left(\frac{\omega}{8}\right)^2 + j0.4\left(\frac{\omega}{8}\right) + 1\right)}$$

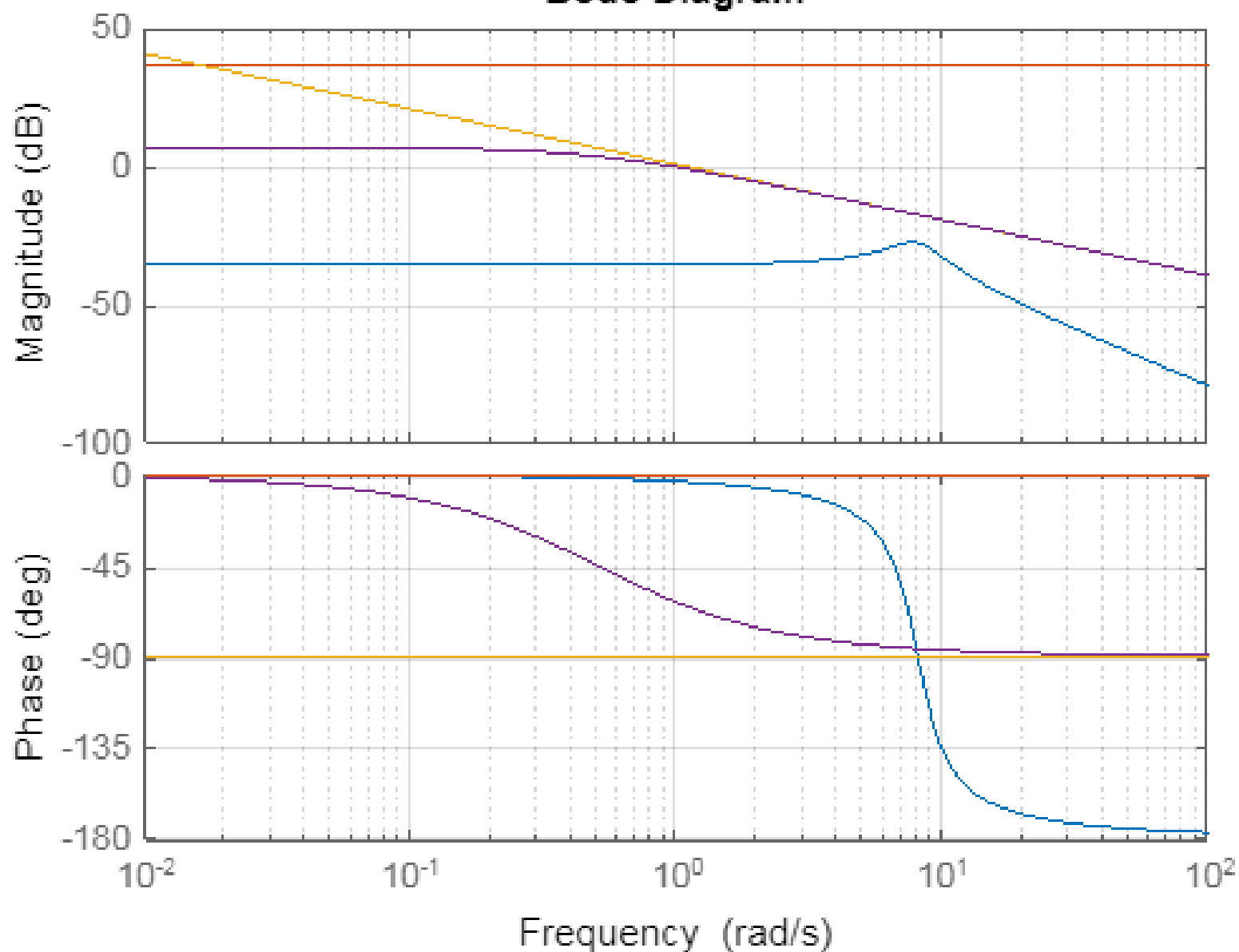
Factors

The factors of this transfer function in order of their increasing frequencies are

1. Constant gain , $k=4$
2. Pole at origin, $1/j\omega$
3. Pole at $s=-0.5$; corner frequency $\omega_1=0.5$
4. zero at $s=-2$; corner frequency $\omega_2=2$
5. Pair of complex conjugate poles with $\zeta =0.2$, $\omega_n=8$; corner frequency $\omega_3 =8$

| Factor | Corner Frequency (rad/sec) | Magnitude Characteristics | Phase Characteristics |
|---|---------------------------------|---|--|
| 4 | - | Straight line of 12 db | zero |
| $\left(\frac{1}{j\omega}\right)$ | - | Straight line of constant slope = -20db/decade passing through zero db at $\omega = 1$ | Constant = -90° |
| $\left(\frac{1}{1 + j 2\omega}\right)$ | $\omega_1=0.5$ | Straight line of 0 db for $\omega < \omega_1$, straight line of slope = -20db/decade for $\omega > \omega_1$ | Varies from 0 to -90°, at $\omega_1 = -45^\circ$ |
| $1 + j0.5\omega$ | $\omega_2 = 2$ | Straight line of 0 db for $\omega < \omega_2$, straight line of slope = 20db/decade for $\omega > \omega_2$ | Varies from 0 to 90°, at $\omega_2 = 45^\circ$ |
| $\frac{1}{1 + j0.4\left(\frac{\omega}{8}\right) - \left(\frac{\omega}{8}\right)^2}$ | $\omega_3 = 8$ $\zeta = 0.2$ | Straight line of 0 db for $\omega < \omega_3$, straight line of slope = -40db/decade for $\omega > \omega_3$ | Varies from 0 to -180°, at $\omega_3 = -90^\circ$ |

Bode Diagram



Problem

The system has an open loop transfer function $G(s) = \frac{10k}{s(1+0.05s)(1+0.1s)}$. Find the gain k such that

(i) $GM = 20\text{dB}$

(ii) $PM = 10^\circ$

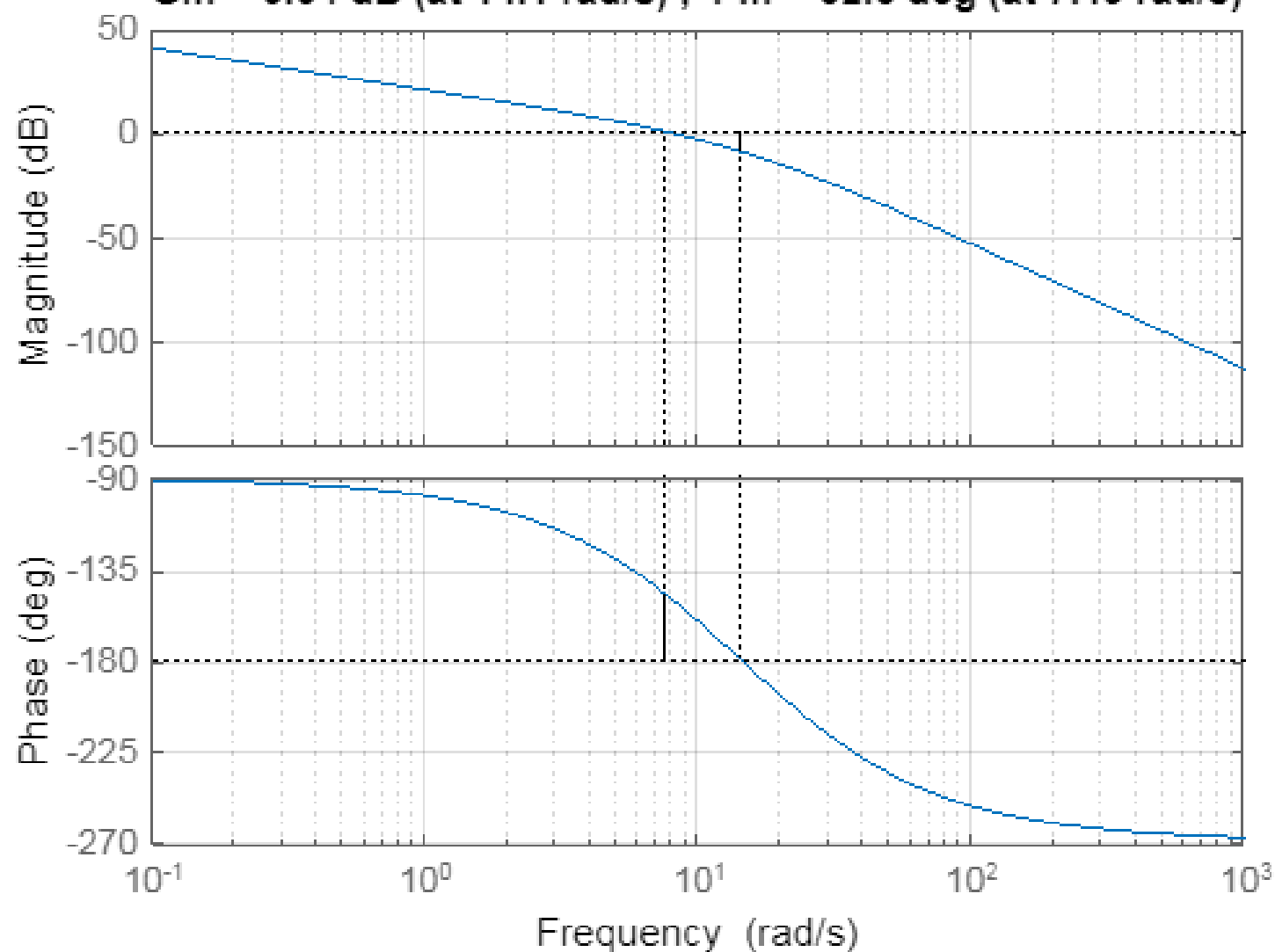
Assume $K=1$, Draw the Bode plot

$$M_{\text{db}} = 20\log 10 - 20\log \omega - 20\log \sqrt{1+(0.05\omega)^2} - 20\log \sqrt{1+(0.1\omega)^2}$$

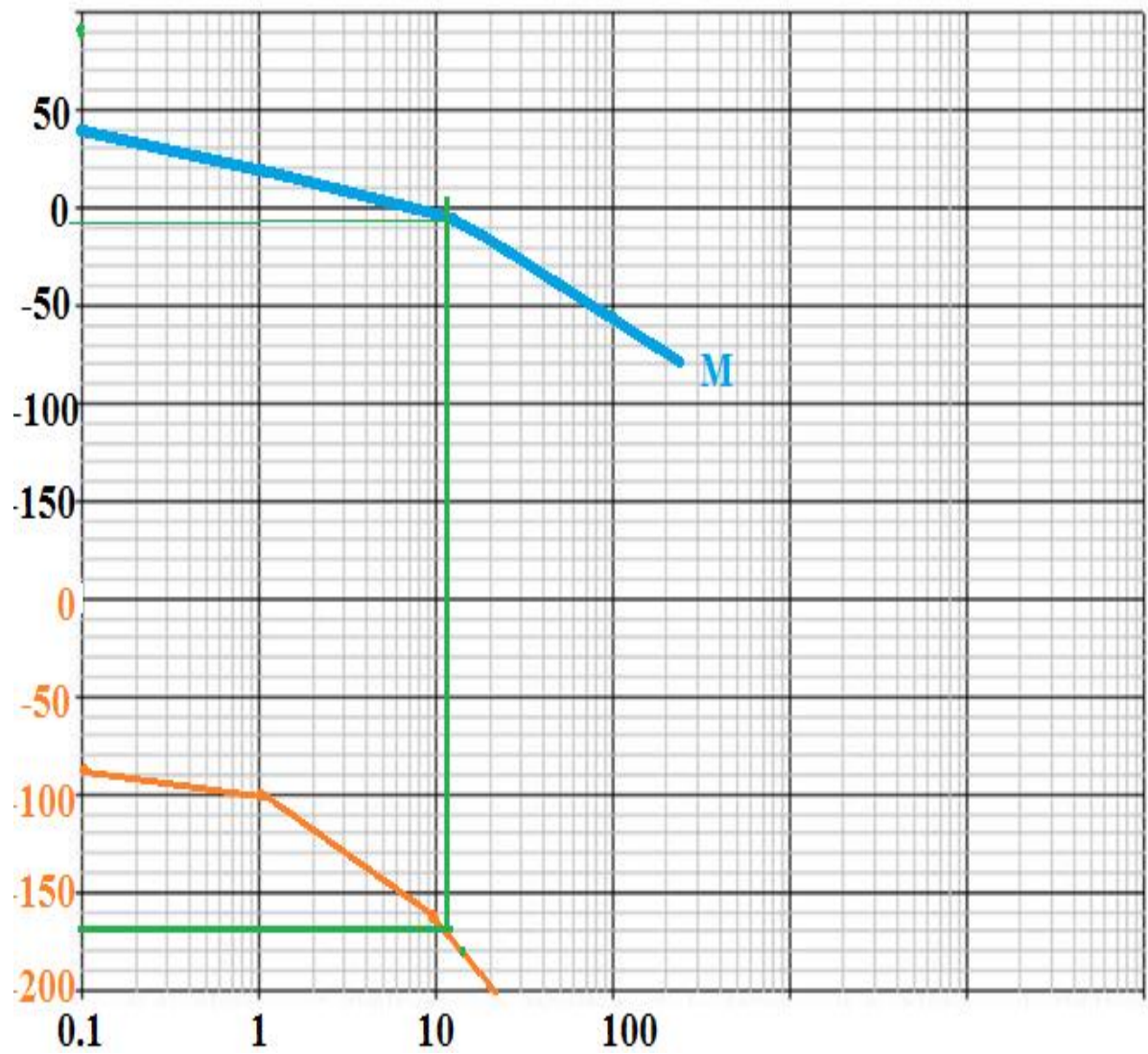
$$\phi = 0 - 90 - \tan^{-1} (0.05\omega) - \tan^{-1} (0.1\omega)$$

Bode Diagram

Gm = 9.54 dB (at 14.1 rad/s) , Pm = 32.6 deg (at 7.49 rad/s)



| ω | M | ϕ |
|----------|-----|--------|
| 0.1 | 40 | -91 |
| 1 | 20 | -99 |
| 10 | -4 | -162 |
| 20 | -16 | -198 |
| 100 | -54 | -253 |



For $k=1$, $\omega_g = 7.49$ rad/sec

$\omega_p = 14.1$ rad/sec

GM = 9.54 dB

PM = 32.6°

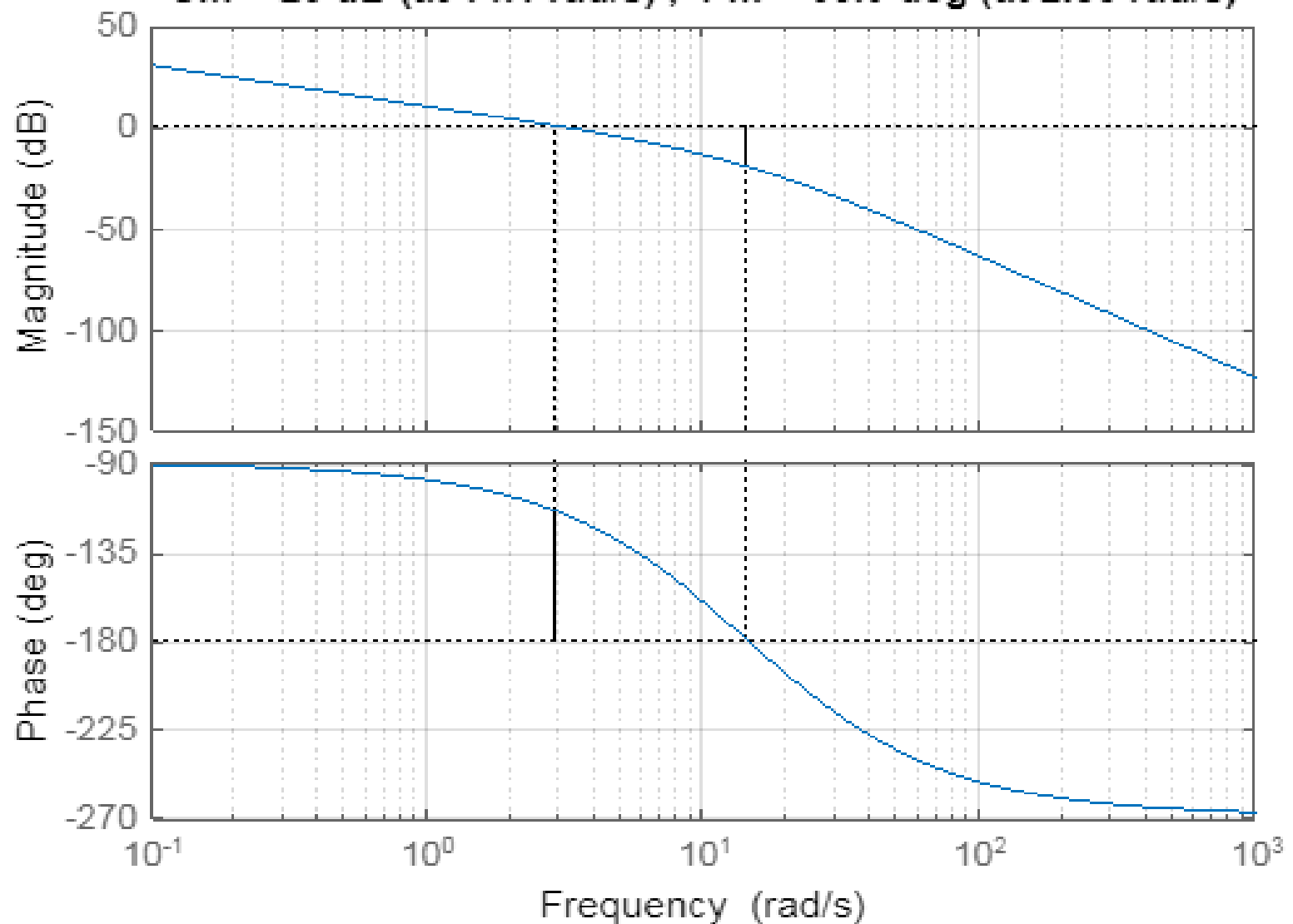
(i) To get GM = 20 dB, the plot is shifted downwards by $(20 - 9.54) = 10.46$ dB.

The system gain is decreased by -10.46 dB or multiplied by a factor 0.32

(ii) To get PM = 10° , the plot is shifted upwards by 7 dB. The system gain is changed by 7 dB or multiplied by a factor 2.2

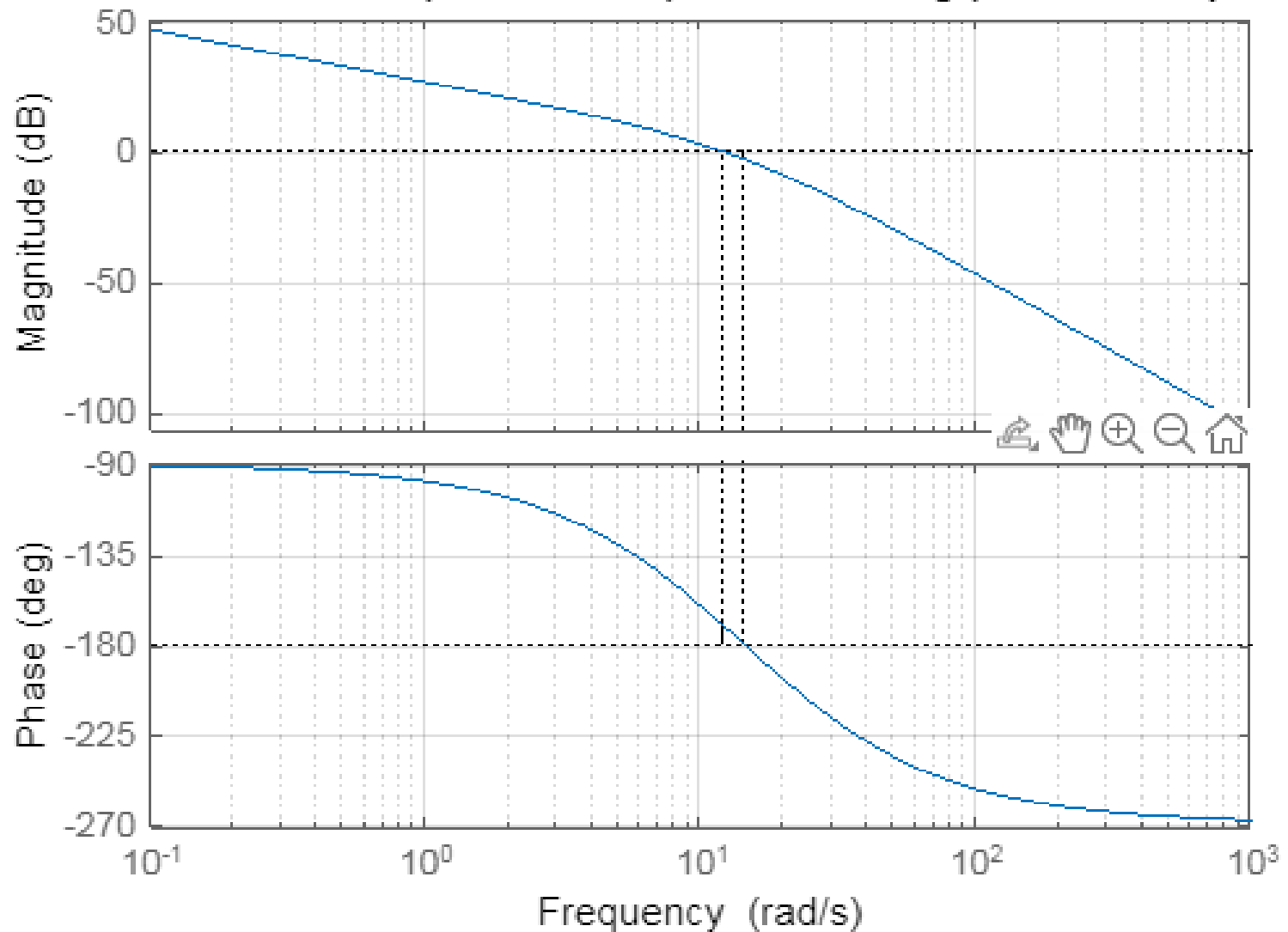
Bode Diagram

Gm = 20 dB (at 14.1 rad/s) , Pm = 65.9 deg (at 2.86 rad/s)



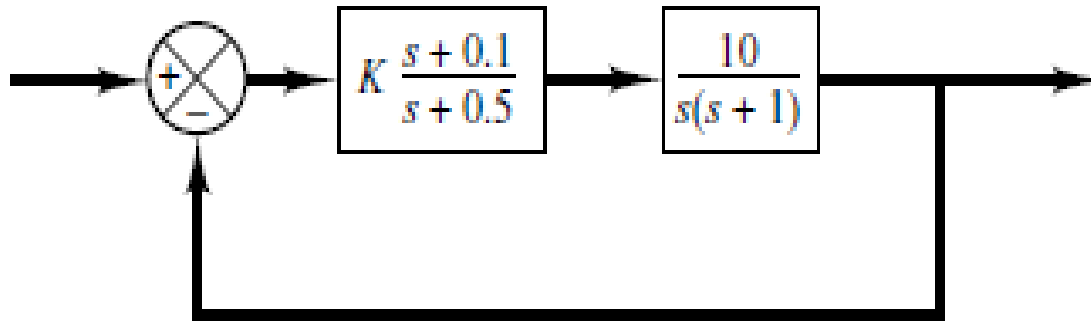
Bode Diagram

$G_m = 3.1 \text{ dB}$ (at 14.1 rad/s) , $P_m = 10 \text{ deg}$ (at 11.7 rad/s)

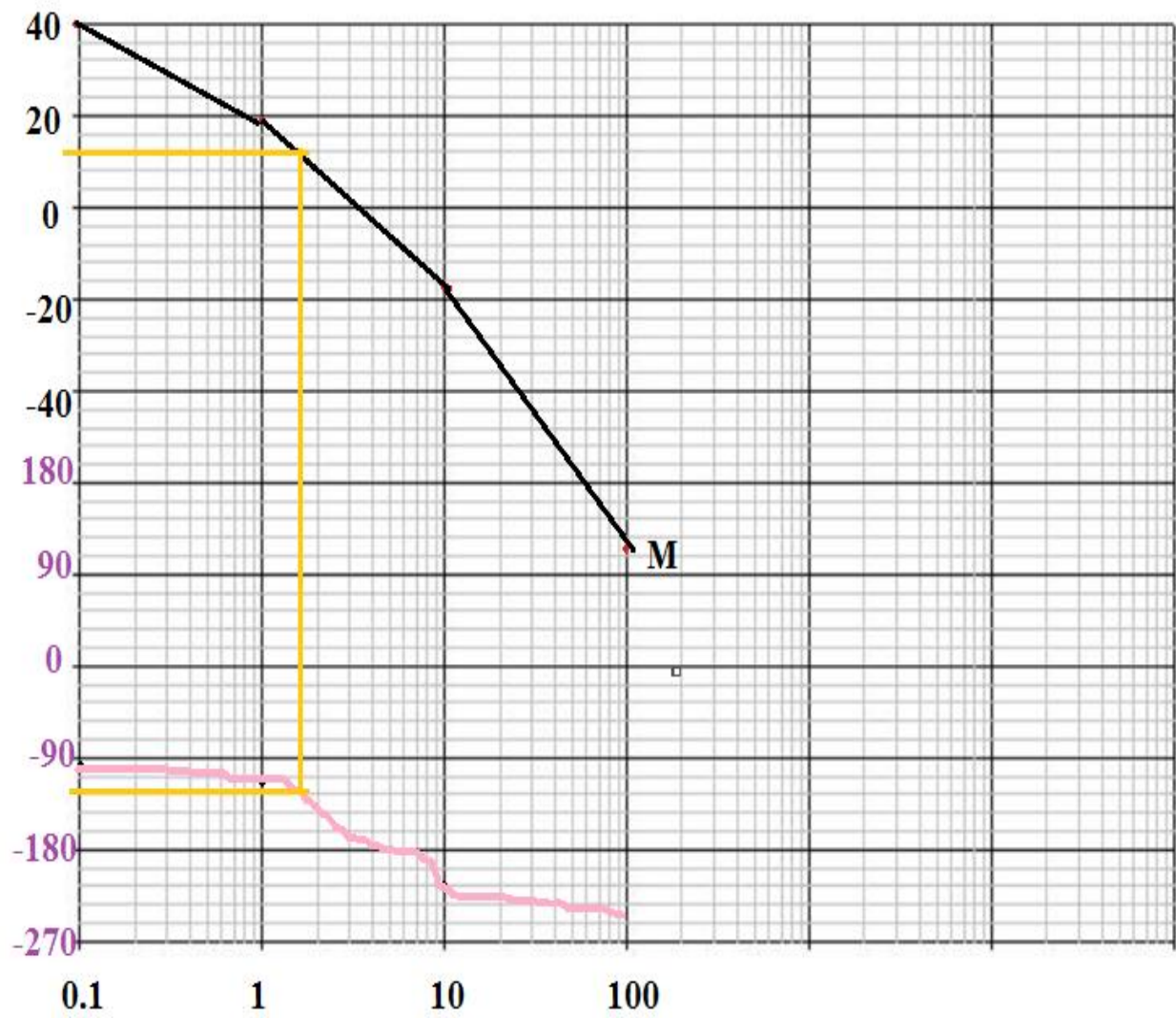


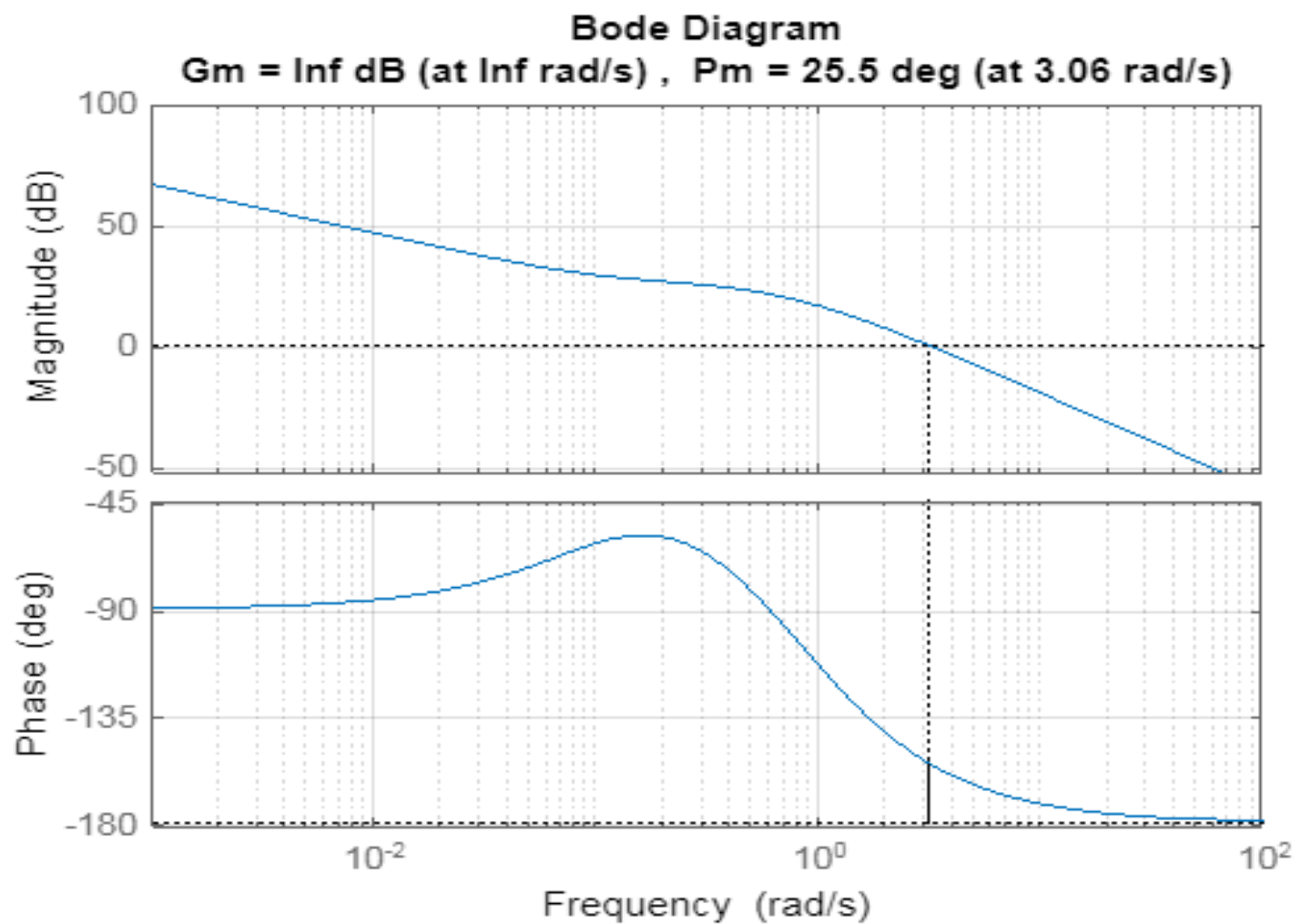
Problem

Consider the system shown in Figure. Draw a Bode diagram of the open-loop transfer function, and determine the value of the gain K such that the phase margin is 50° . What is the gain margin of this system with this gain K ?

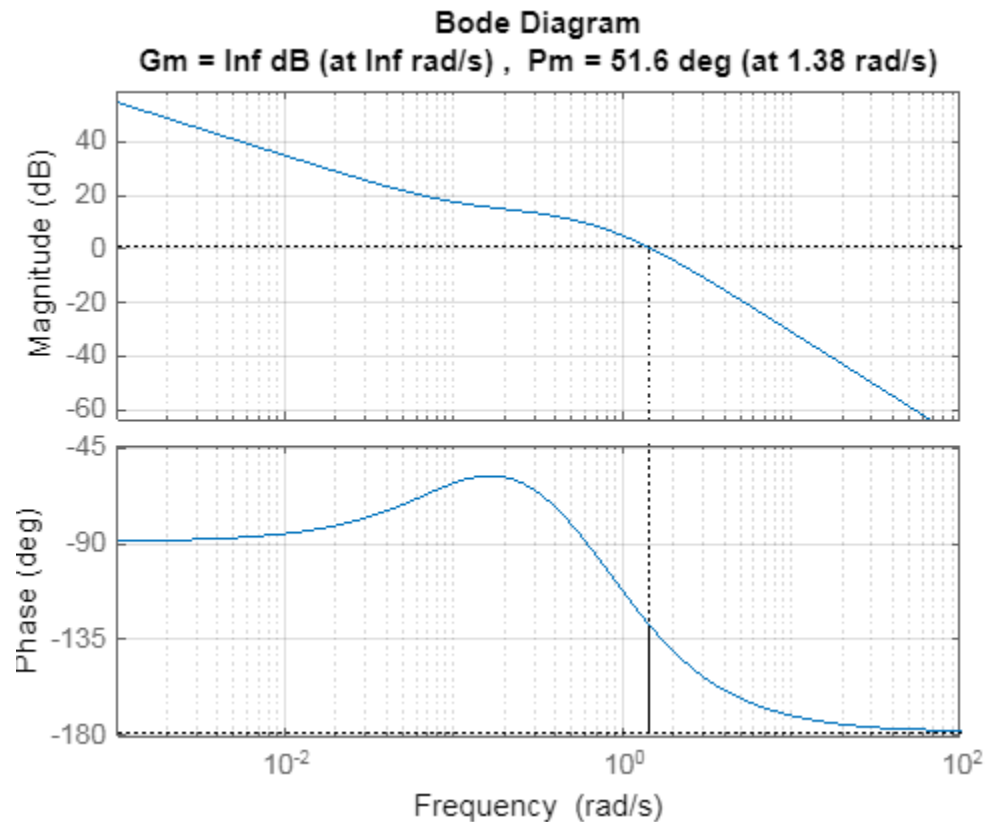


| ω | M (db) | ϕ (deg) |
|----------|--------|--------------|
| 0.1 | 40 | -93 |
| 1 | 19 | -122.3 |
| 2 | 11 | -146 |
| 10 | -17.16 | -213.7 |
| 100 | -74 | -263 |

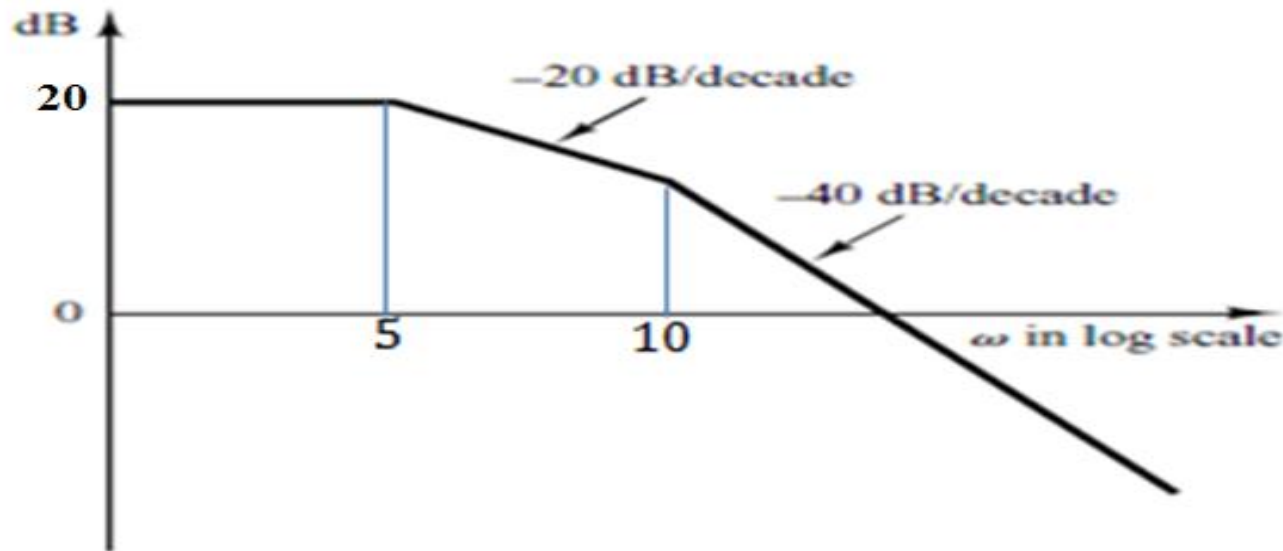




To get $PM = 50^\circ$, the plot is shifted upwards by 8 dB. The system gain is changed by 8 dB or increased by a factor 2.5



Transfer Function from Bode Plot



$$G(s) = \frac{k}{\left(1+\frac{s}{5}\right)\left(1+\frac{s}{10}\right)} = \frac{k}{\left(\frac{5+s}{5}\right)\left(\frac{10+s}{10}\right)} = \frac{k*50}{(5+s)(10+s)}$$

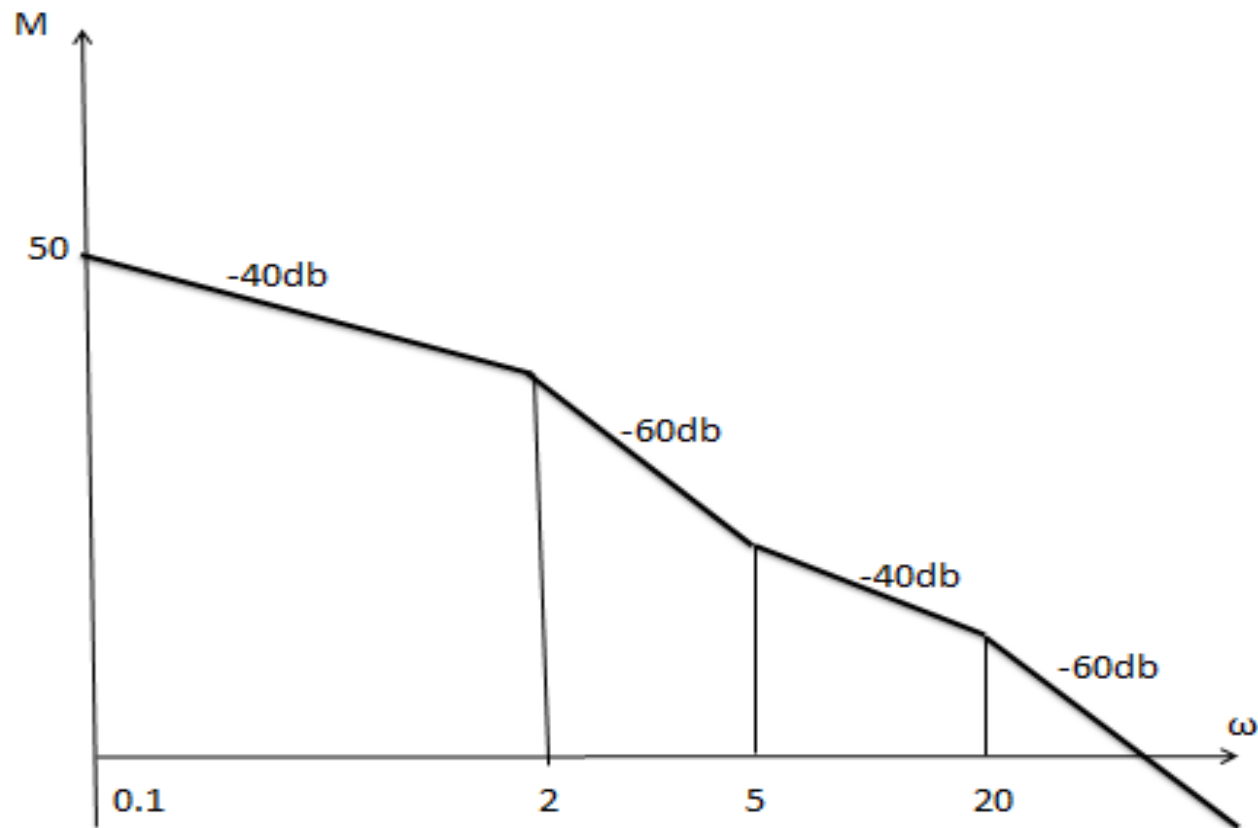
$$20 \log k = 20$$

$$k=1$$

$$G(s) = \frac{50}{(5+s)(10+s)}$$

Problem

Find the transfer function for the given Bode plot



$$G(s) = \frac{k(1+\frac{s}{5})}{s^2(1+\frac{s}{2})(1+\frac{s}{20})}$$

At $\omega = 0.1$, $M = 50\text{db}$

$$20 \log(\frac{k}{\omega^2}) = 50$$

$$20 \log k + 20 \log(\frac{1}{\omega^2}) = 50$$

$$20 \log k - 40 \log \omega = 50$$

$$20 \log k - 40 \log (0.1) = 50$$

$$20 \log k + 40 = 50$$

$$20 \log k = 10, K = 10^{10/20} = 3.16$$

$$G(s) = \frac{3.16(\frac{5+s}{5})}{s^2(\frac{2+s}{2})(\frac{20+s}{20})} = \frac{25.28(s+5)}{s^2(2+s)(20+s)}$$

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